

Antennas for Wireless Body Area Networks

The 24th International Traveling Summer School
on Microwaves and Lightwaves

Zbynek Raida

OUTLINE

- ❑ **Introduction to antennas**
From Maxwell equations to applications
- ❑ **Human body**
Ways of modeling
- ❑ **Antennas**
On-body versus off-body
- ❑ **Conclusions**

Maxwell equations (1)



James Clerk Maxwell

- **1864**: The agreement of the results seems to show that light and magnetism are affections of the same substance, and that light is an electromagnetic disturbance propagated through the field according to electromagnetic laws.

Maxwell equations (2)

$$\oint_l \mathbf{H} \cdot d\mathbf{l} = I + \frac{d}{dt} \int_s (\boldsymbol{\varepsilon} \mathbf{E}) \cdot d\mathbf{S}$$

$$\nabla \times \mathbf{H} = \mathbf{J} + \frac{d}{dt} (\boldsymbol{\varepsilon} \mathbf{E})$$

$$\oint_l \mathbf{E} \cdot d\mathbf{l} = -\frac{d}{dt} \int_s (\boldsymbol{\mu} \mathbf{H}) \cdot d\mathbf{S}$$

$$\nabla \times \mathbf{E} = -\frac{d}{dt} (\boldsymbol{\mu} \mathbf{H})$$

$$\oint_s (\boldsymbol{\varepsilon} \mathbf{E}) \cdot d\mathbf{S} = Q$$

$$\nabla \cdot (\boldsymbol{\varepsilon} \mathbf{E}) = \rho$$

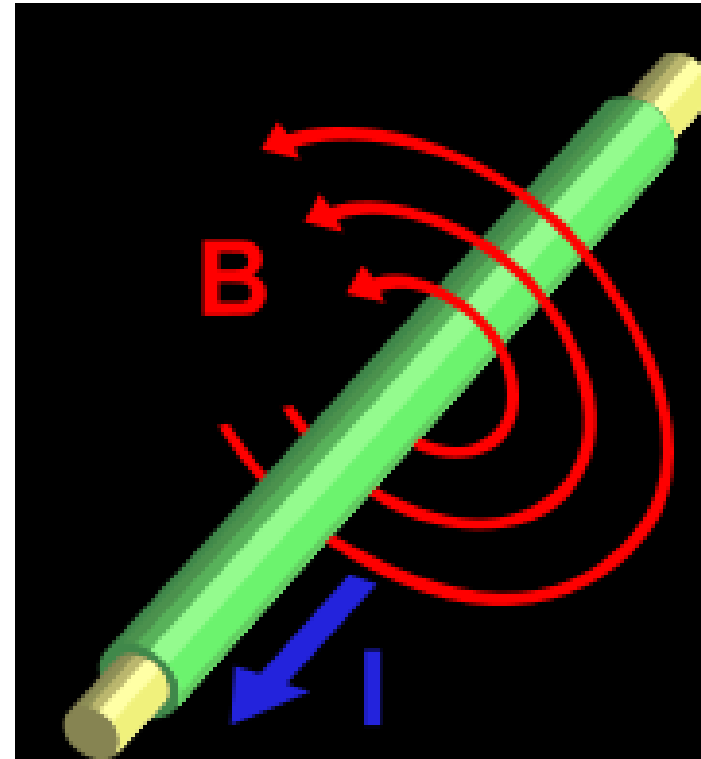
$$\oint_s (\boldsymbol{\mu} \mathbf{H}) \cdot d\mathbf{S} = 0$$

$$\nabla \cdot (\boldsymbol{\mu} \mathbf{H}) = 0$$

Maxwell equations (3)



André-Marie Ampère

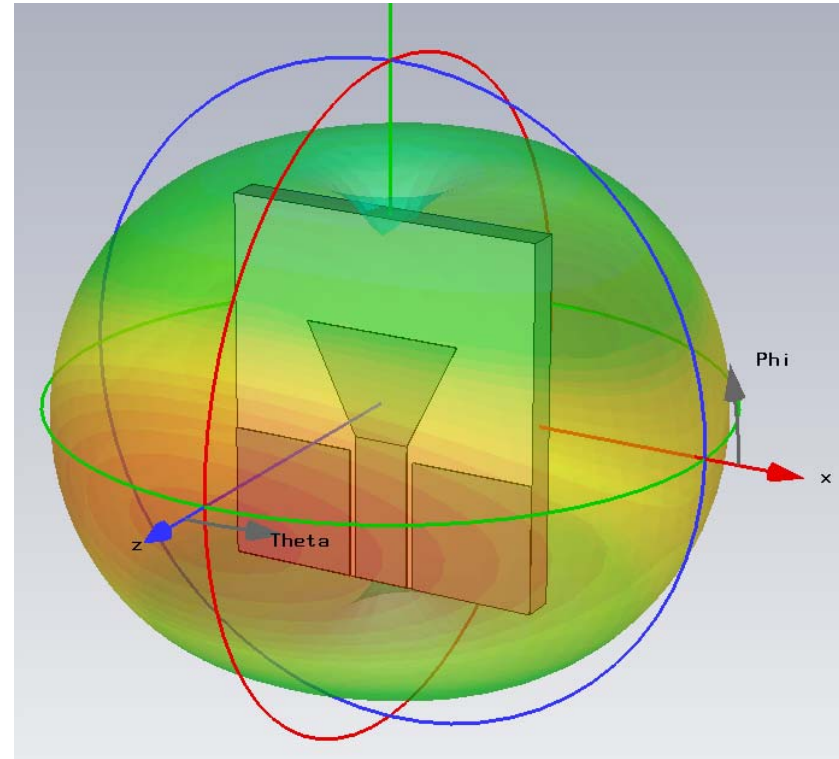


$$\oint_l \mathbf{H} \cdot d\mathbf{l} = I + \frac{d}{dt} \int_s (\varepsilon \mathbf{E}) \cdot d\mathbf{S}$$

Maxwell equations (4)



André-Marie Ampère

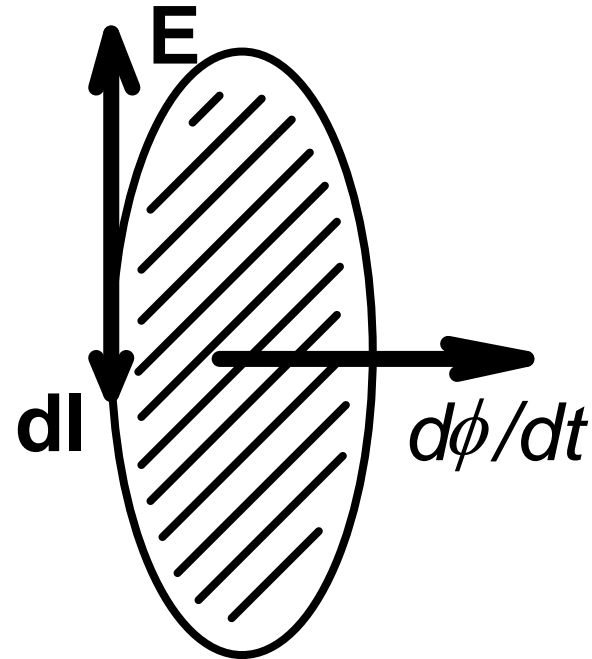


$$\oint_l \mathbf{H} \cdot d\mathbf{l} = I + \frac{d}{dt} \int_s (\varepsilon \mathbf{E}) \cdot d\mathbf{S}$$

Maxwell equations (5)

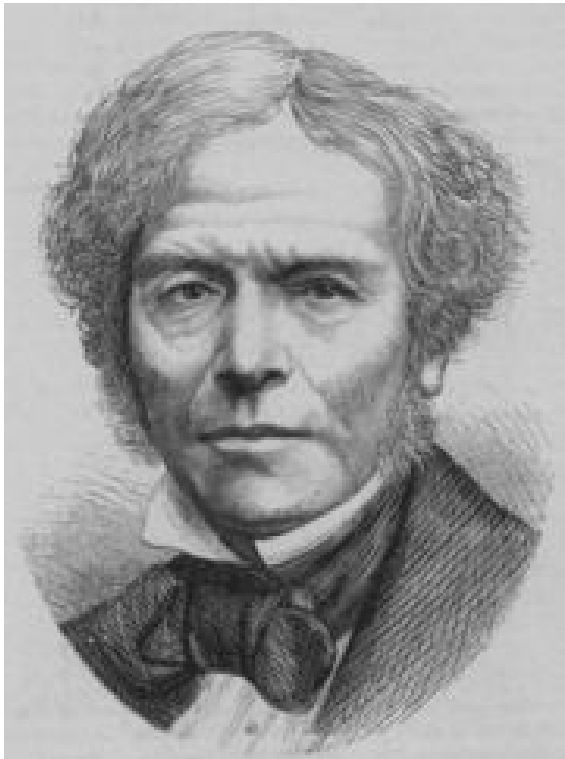


Michael Faraday

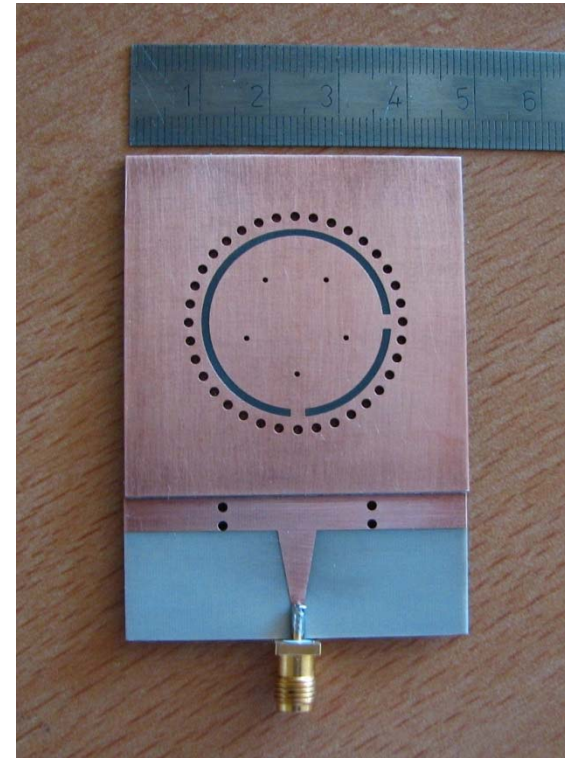


$$\oint_l \mathbf{E} \cdot d\mathbf{l} = -\frac{d}{dt} \int_s (\mu \mathbf{H}) \cdot d\mathbf{S}$$

Maxwell equations (6)



Michael Faraday



$$\oint_l \mathbf{E} \cdot d\mathbf{l} = -\frac{d}{dt} \int_s (\mu \mathbf{H}) \cdot d\mathbf{S}$$

Wave equation

- Let's assume harmonic fields
- Then, Maxwell equations

$$\nabla \times \mathbf{E} = -j\omega\mu \mathbf{H}$$

$$\nabla \times \mathbf{H} = \mathbf{J} + j\omega\varepsilon \mathbf{E}$$

- Reducing the number of unknowns → potentials

$$\mathbf{B} = \nabla \times \mathbf{A} \quad \nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J} \quad k^2 = \omega^2 \varepsilon \mu$$

$$\mathbf{E} + j\omega\mu \mathbf{A} = -\nabla \varphi \quad \nabla^2 \varphi + k^2 \varphi = 0$$

$$\varphi = \left(j\omega/k^2 \right) \nabla \cdot \mathbf{A}$$

Potentials

- Currents flowing along antenna wire

$$\mathbf{A}(\mathbf{r}) = \mu_0 \iiint \mathbf{J}(\mathbf{r}') \frac{\exp[-j \mathbf{k} \cdot (\mathbf{r} - \mathbf{r}')] }{4\pi |\mathbf{r} - \mathbf{r}'|} d\mathbf{r}'$$

- Charges accumulating at the ends of antenna

$$\varphi(\mathbf{r}) = \frac{1}{\varepsilon_0} \iiint q(\mathbf{r}') \frac{\exp[-j \mathbf{k} \cdot (\mathbf{r} - \mathbf{r}')] }{4\pi |\mathbf{r} - \mathbf{r}'|} d\mathbf{r}'$$

- Continuity theorem

$$-j \omega q(\mathbf{r}) = \nabla \cdot \mathbf{J}$$

- Radiated wave

$$\mathbf{E}(\mathbf{r}) = -j \omega \mathbf{A}(\mathbf{r}) - \nabla \varphi$$

Elementary dipole (1)

- Magnetic vector potential

$$\mathbf{A}(\mathbf{r}) = \mu_0 \iiint \mathbf{J}(\mathbf{r}') \frac{\exp[-j\mathbf{k} \cdot (\mathbf{r} - \mathbf{r}')] }{4\pi |\mathbf{r} - \mathbf{r}'|} d\mathbf{r}'$$

$$dA_z = \frac{\mu_0}{4\pi} I dz \frac{e^{-jkr}}{r}$$

Elementary dipole (2)

- Substituting vector potential to Maxwell equations

$$dE_r = \frac{1}{4\pi\epsilon} \frac{2I dz}{\omega} k^3 \cos \vartheta \left[\frac{-j}{(kr)^3} + \frac{1}{(kr)^2} \right] e^{-jkr}$$

$$dE_\vartheta = \frac{1}{4\pi\epsilon} \frac{I dz}{\omega} k^3 \sin \vartheta \left[\frac{-j}{(kr)^3} + \frac{1}{(kr)^2} + \frac{j}{(kr)} \right] e^{-jkr}$$

$$dH_\varphi = \frac{1}{4\pi} I dz k^2 \sin \vartheta \left[\frac{1}{(kr)^2} + \frac{j}{(kr)} \right] e^{-jkr}$$

Elementary dipole (3)

- Far field

$$dE_{\vartheta} = 60I j \frac{k}{2} \sin \vartheta dz \frac{e^{-jkr}}{r}$$

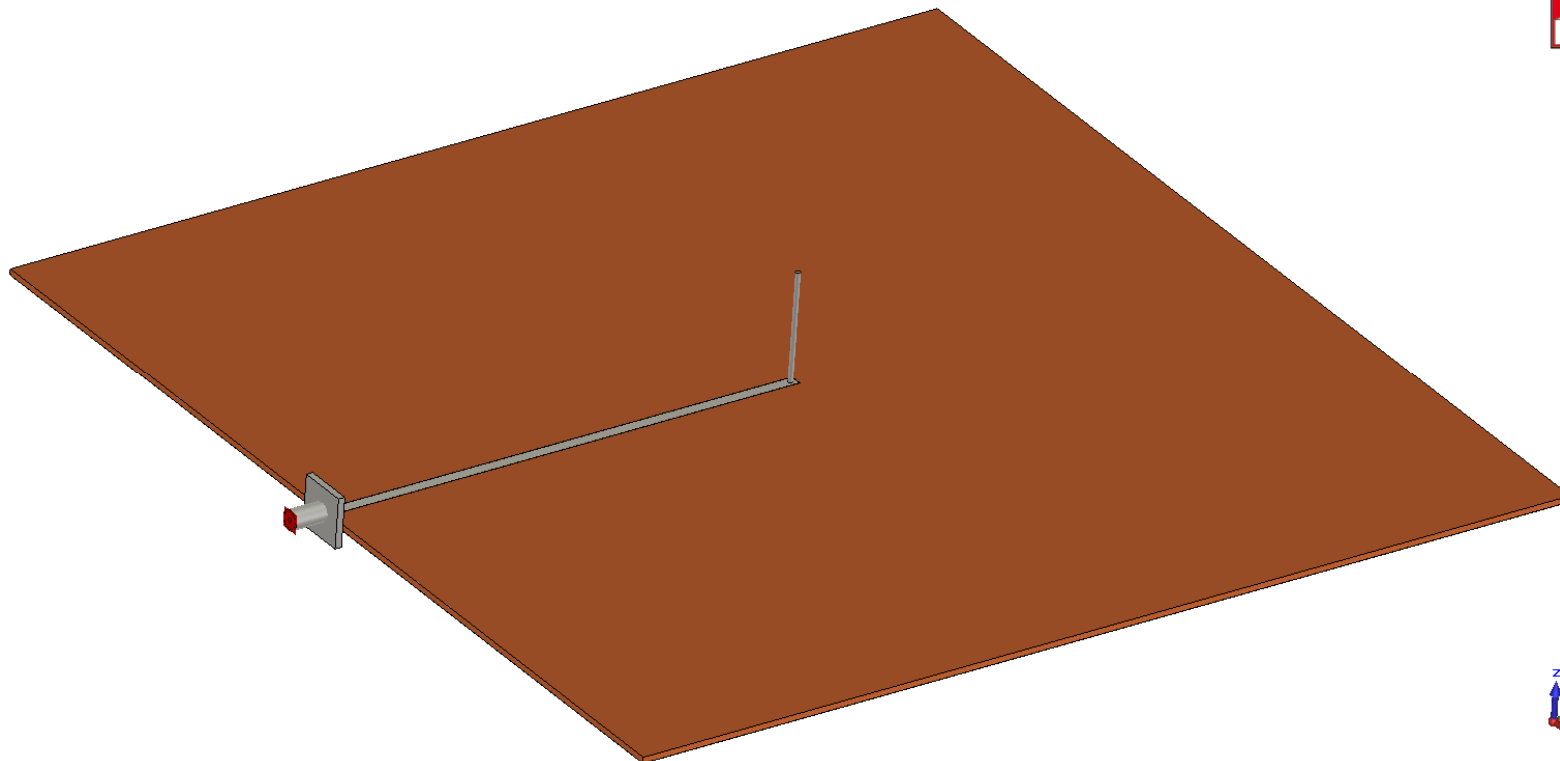
$$dH_{\varphi} = \frac{dE_{\vartheta}}{120\pi}$$

- Radiation pattern

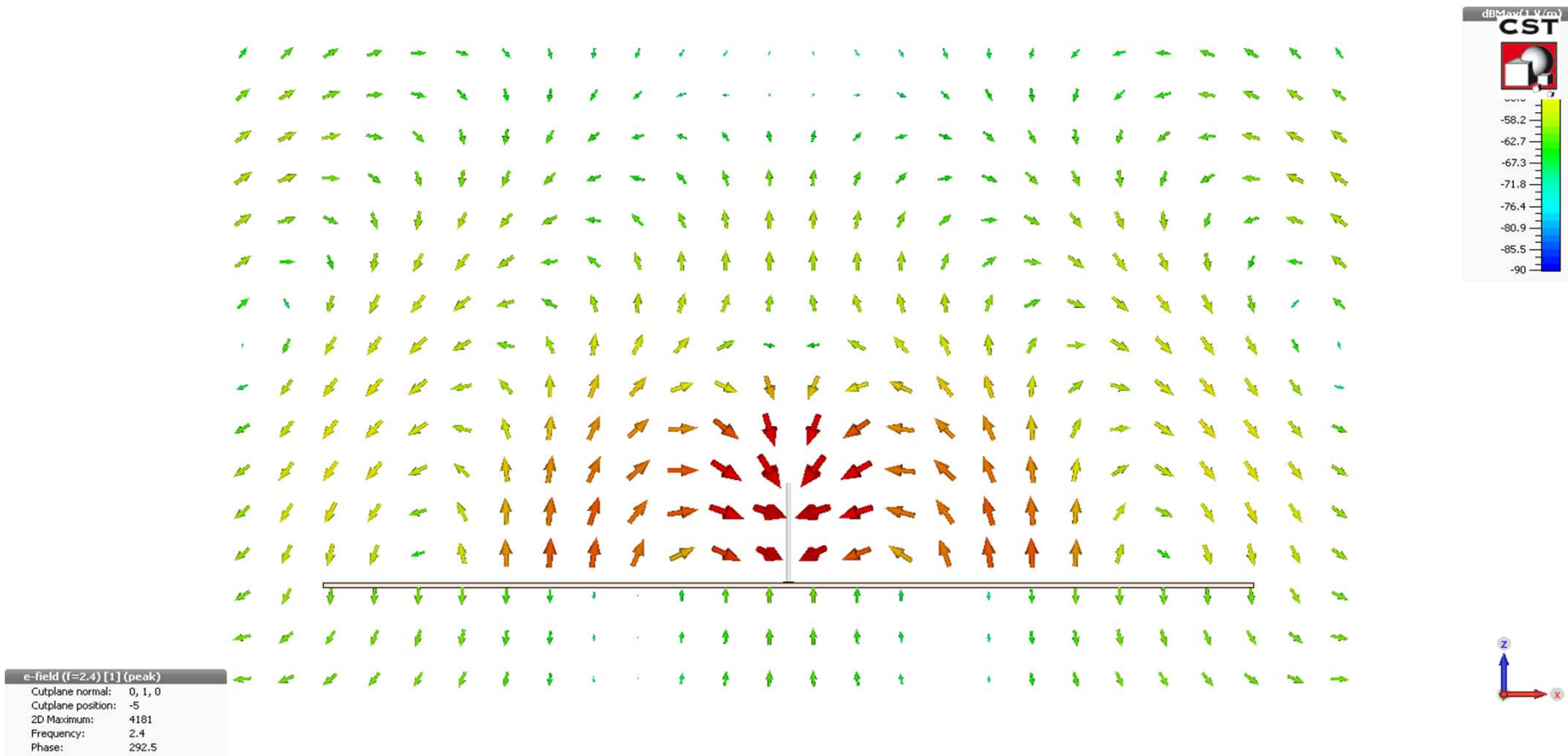
$$E = 60I F(\varphi, \vartheta) \frac{e^{-jkr}}{r}$$

$$F(\varphi, \vartheta) = j \frac{k}{2} \sin \vartheta dz$$

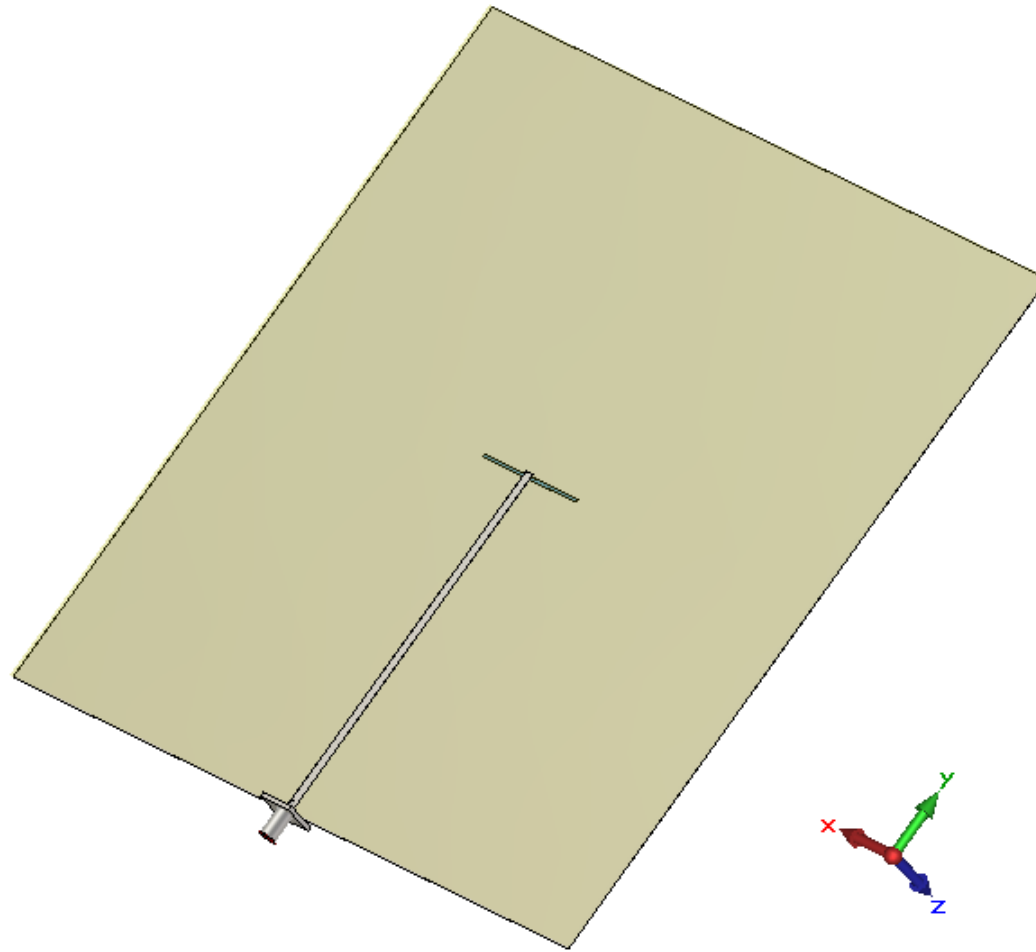
Elementary dipole (4)



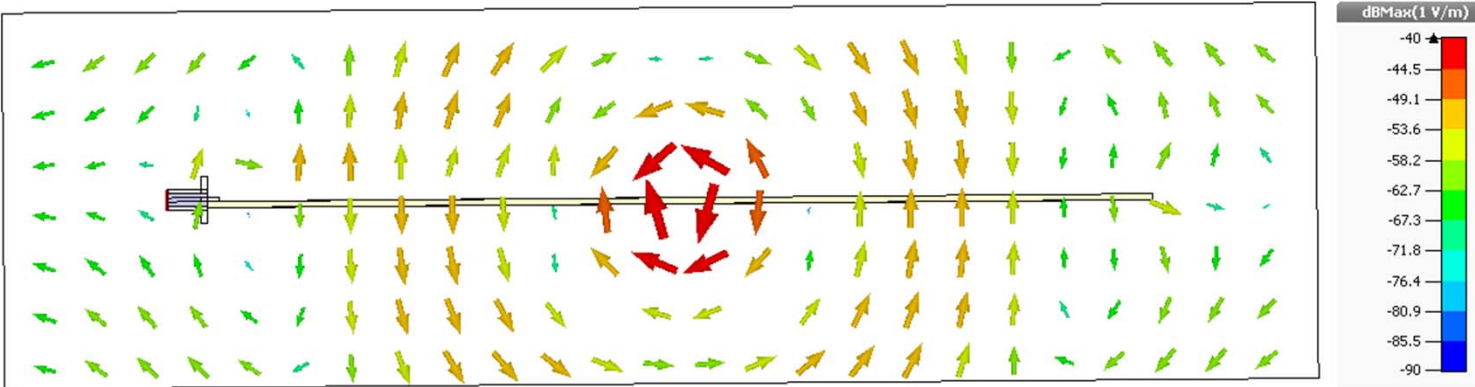
Elementary dipole (5)



Elementary dipole (6)



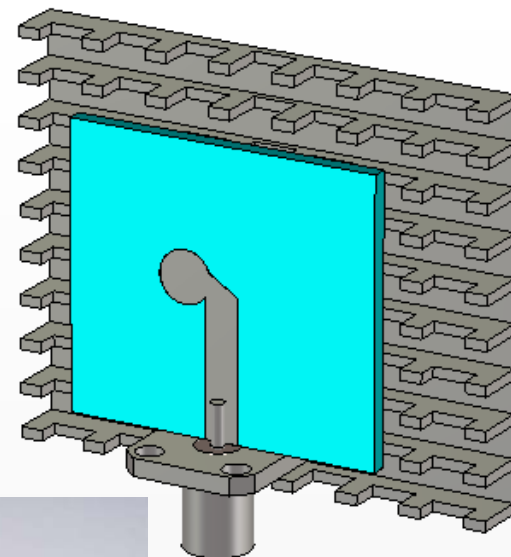
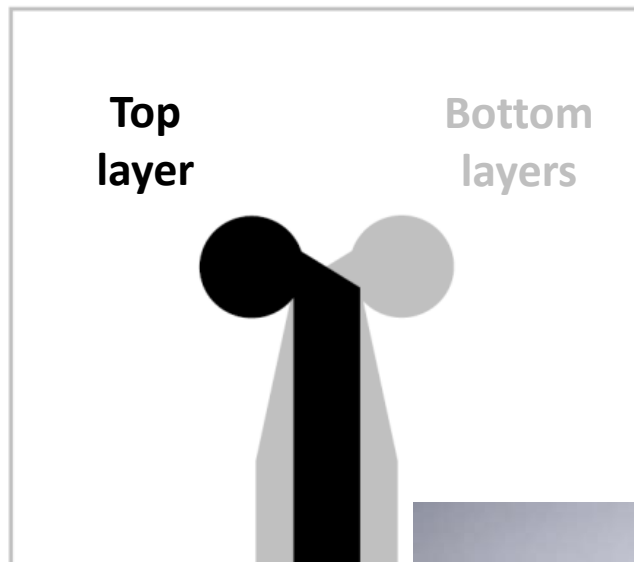
Elementary dipole (7)



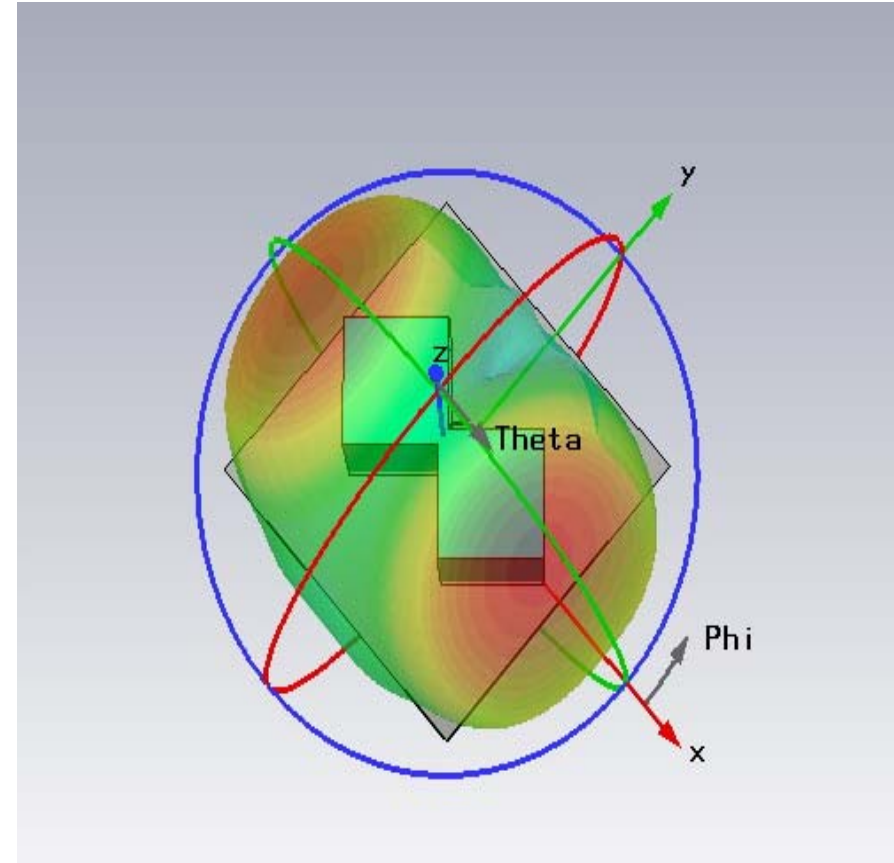
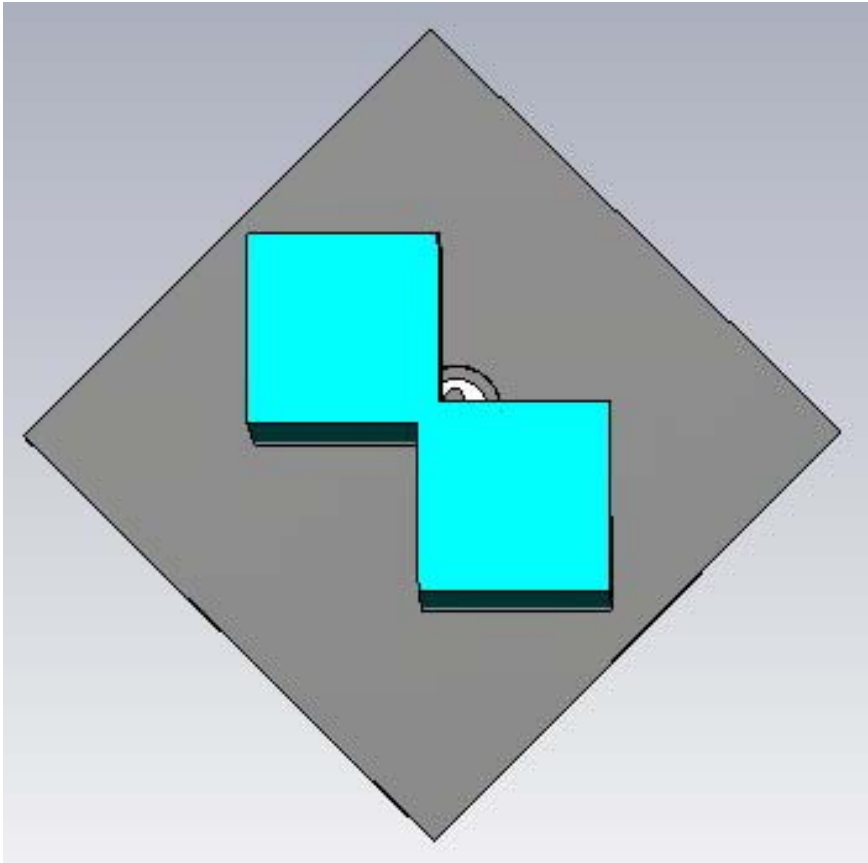
e-field (f=2.4) [1] (peak)	
Cutplane normal:	1, 0, 0
Cutplane position:	0
2D Maximum:	3.386e+04
Frequency:	2.4
Phase:	22.5



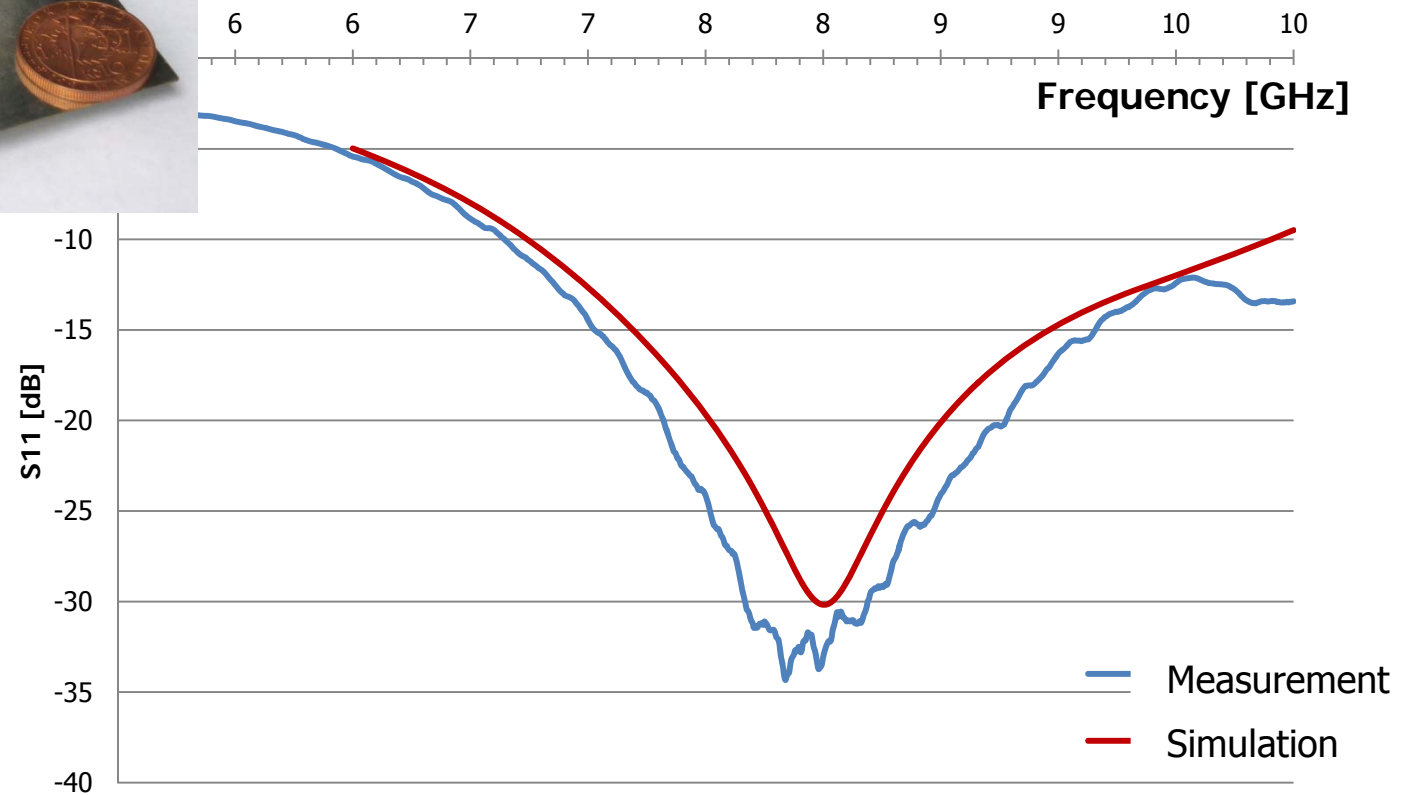
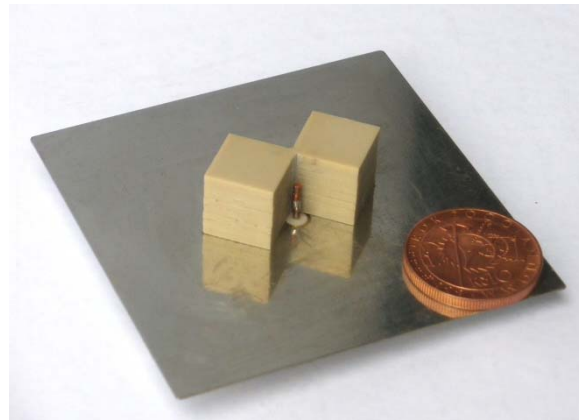
Practical dipoles (1)



Practical dipoles (2)



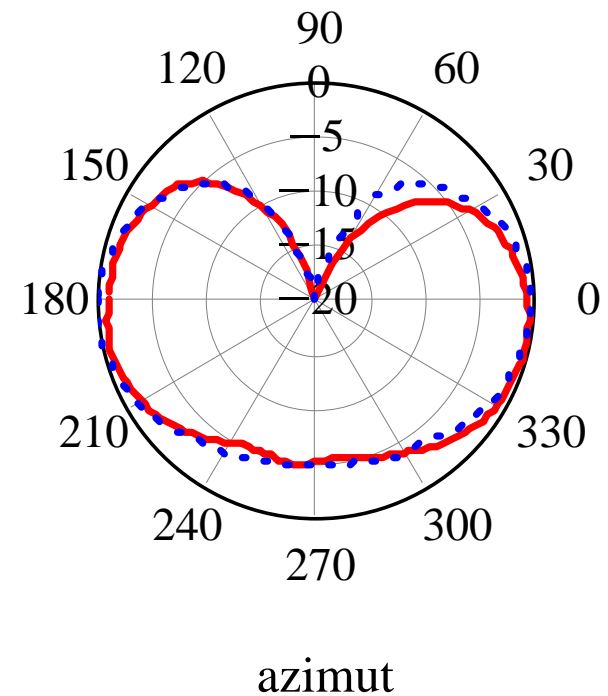
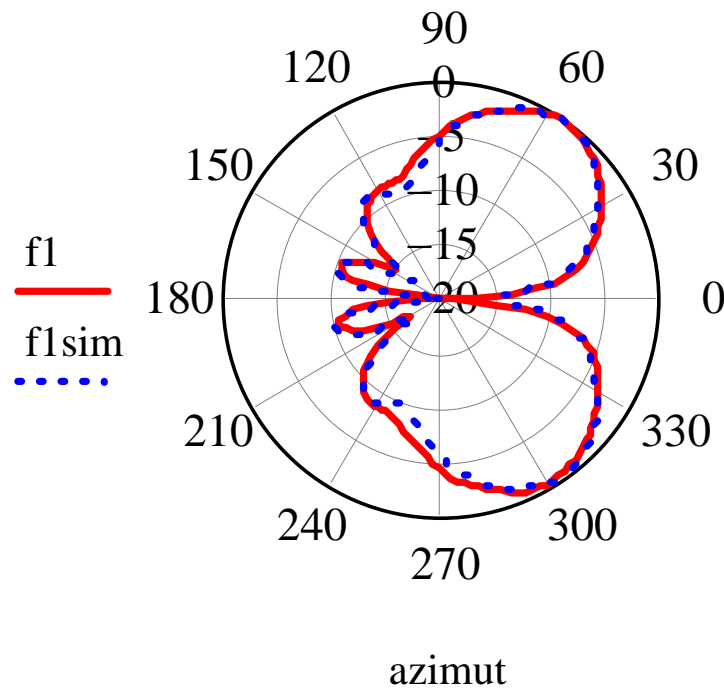
Practical dipoles (3)



Practical dipoles (4)

$f = 7.30$ GHz **E plane**

H plane



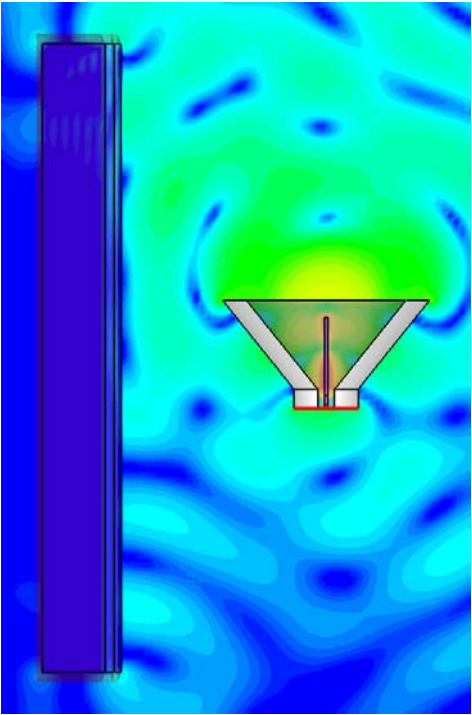
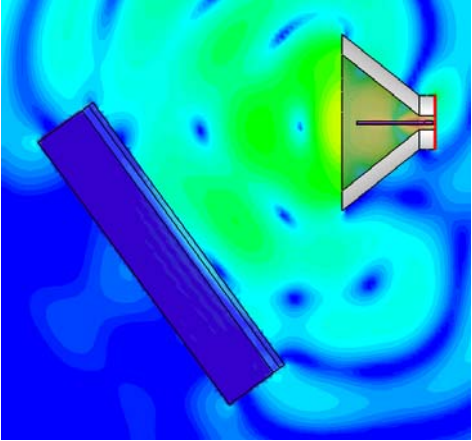
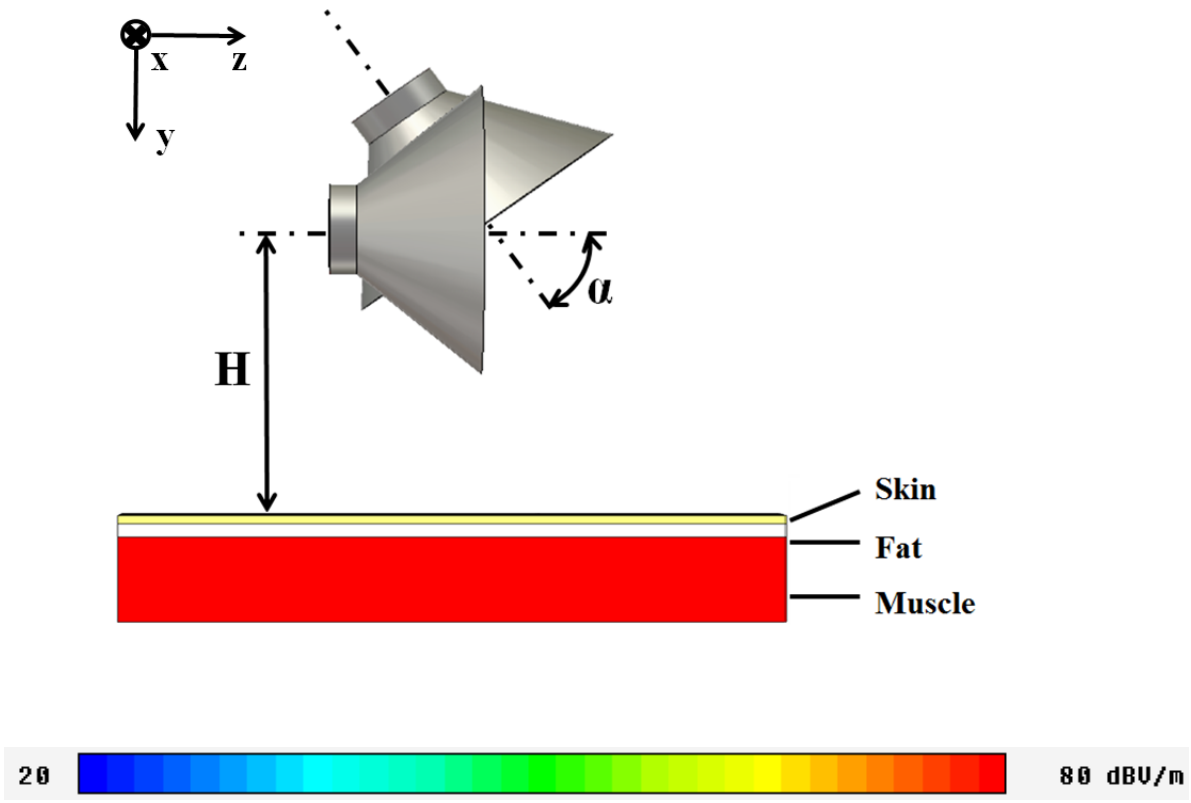
Towards tissue

- Surface and creeping waves:
tangential electric field intensity attenuated
- Monopoles exciting vertical polarization
versus slot antennas → in parallel to skin
- Sensors → directional antennas
- Central unit → omnidirectional antenna
- Testing → phantoms

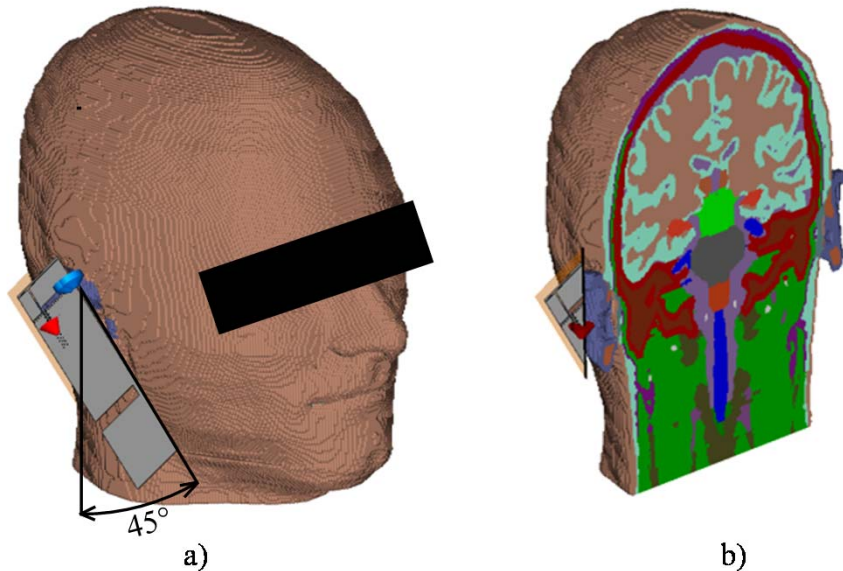
Parameters of tissues

	ϵ_r [-]	ϵ_r [-]	σ [S/m]	σ [S/m]
Tissue	2.40 GHz	5.80 GHz	2.40 GHz	5.80 GHz
Skin	38.063	35.114	1.441	3.717
Fat	5.285	4.955	0.102	0.293
Muscle	52.791	48.485	1.705	4.962

Numerical phantoms (1)

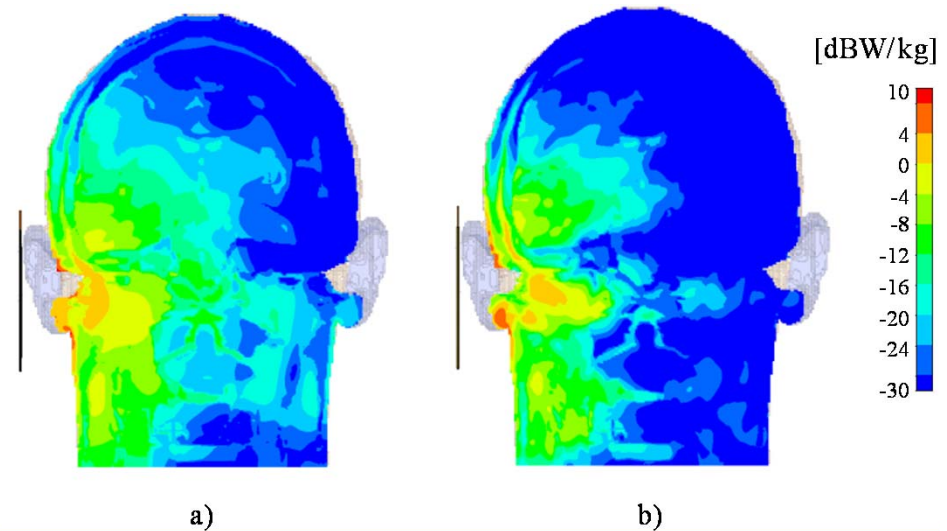


Numerical phantoms (2)



Detailed anatomical models of the human body provided by the ITIS foundation

A. Christ, W. Kainz, E. G. Hahn, K. Honegger, M. Zefferer, E. Neufeld, W. Rascher, R. Janka, W. Bautz, J. Chen, B. Kiefer, P. Schmitt, H. P. Hollenbach, J. X. Shen, M. Oberle, N. Kuster, The virtual family—Development of anatomical CAD models of two adults and two children for dosimetric simulations, *Phys. Med. Biol.*, vol. 55, no. 2, pp. 23-38, Jan. 2010



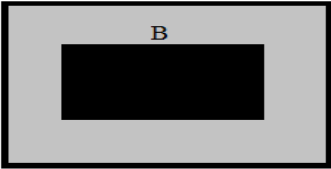
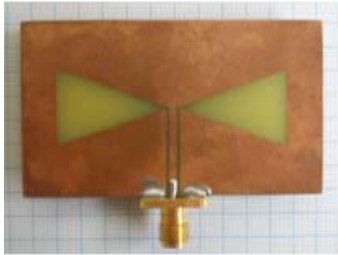
Physical phantoms

- Block of the basis 200 mm × 200 mm and the height 20 mm
- Modified agar gelatin ($\epsilon_r = 48$)
- Agar: 66%
Water: 33%

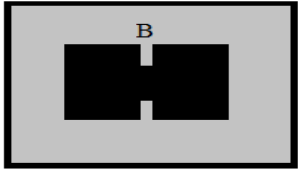
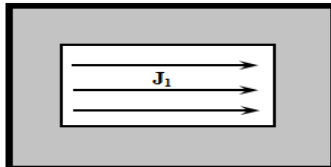


Fractal slot dipole (1)

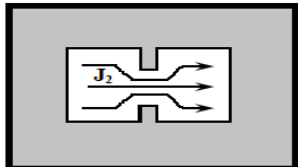
Notches: the same current length in reduced area



$$B \cong \frac{\lambda g}{2}$$



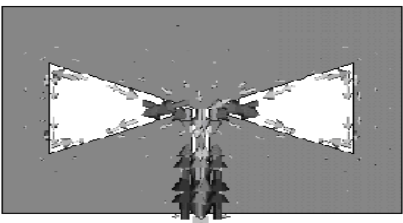
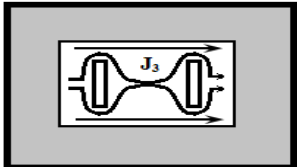
$$B < \frac{\lambda g}{2}$$



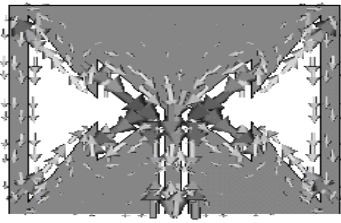
$$J_1 = J_2 = J_3$$



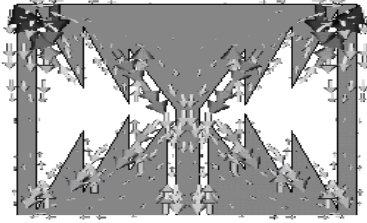
$$B < \frac{\lambda g}{2}$$



Zero Iteration



First Iteration

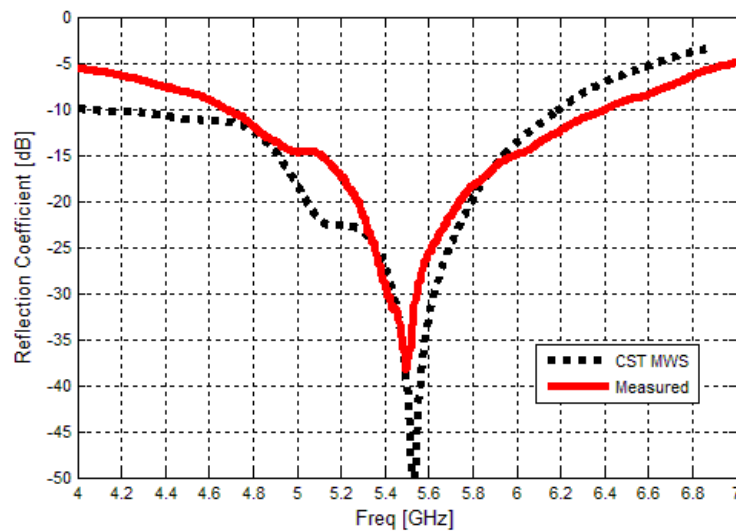


Second Iteration

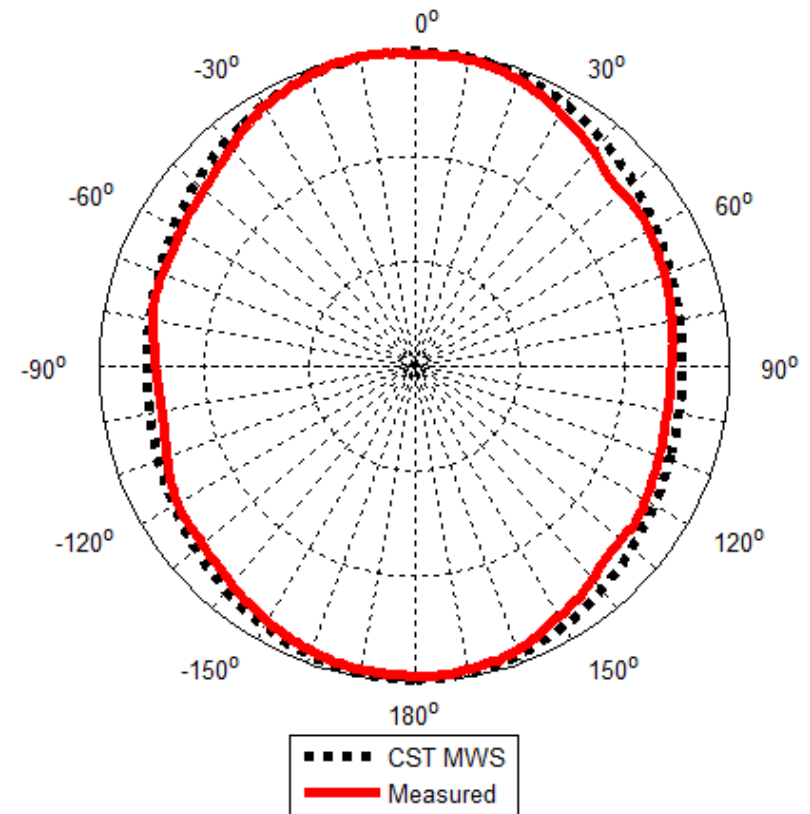
Optimum ratio:

$$1.00 : 0.58 : 0.43$$

Fractal slot dipole (2)



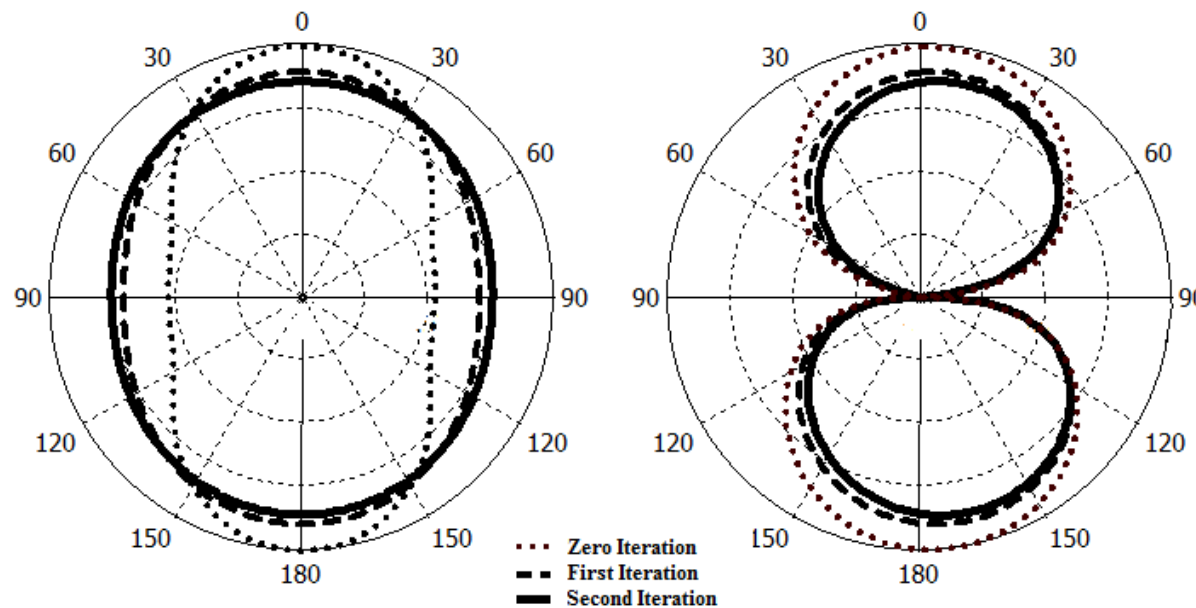
BW = 39.64%



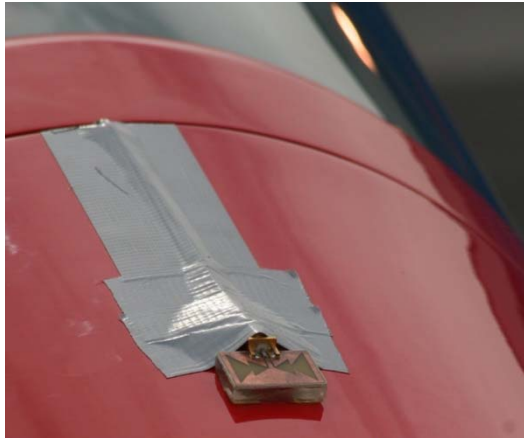
H – Plane
G = 4.1 dBi@5.5GHz

Fractal slot dipole (3)

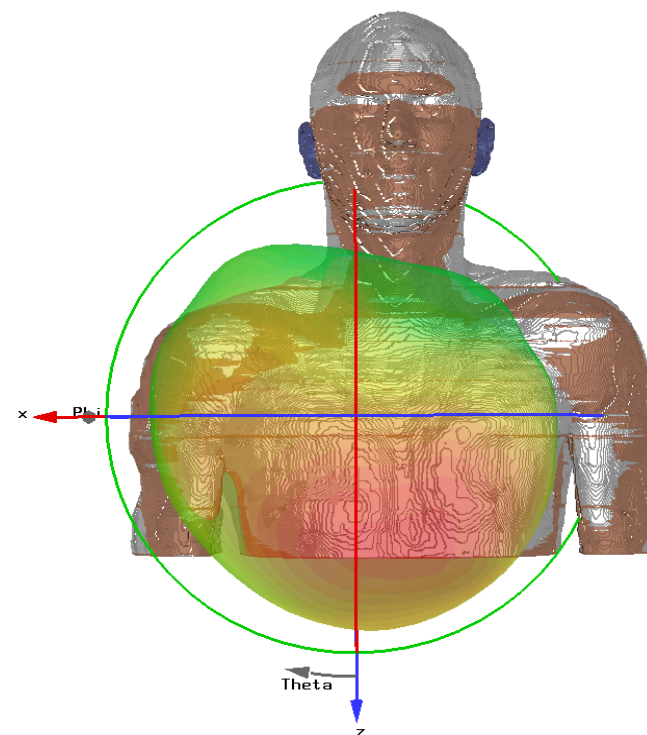
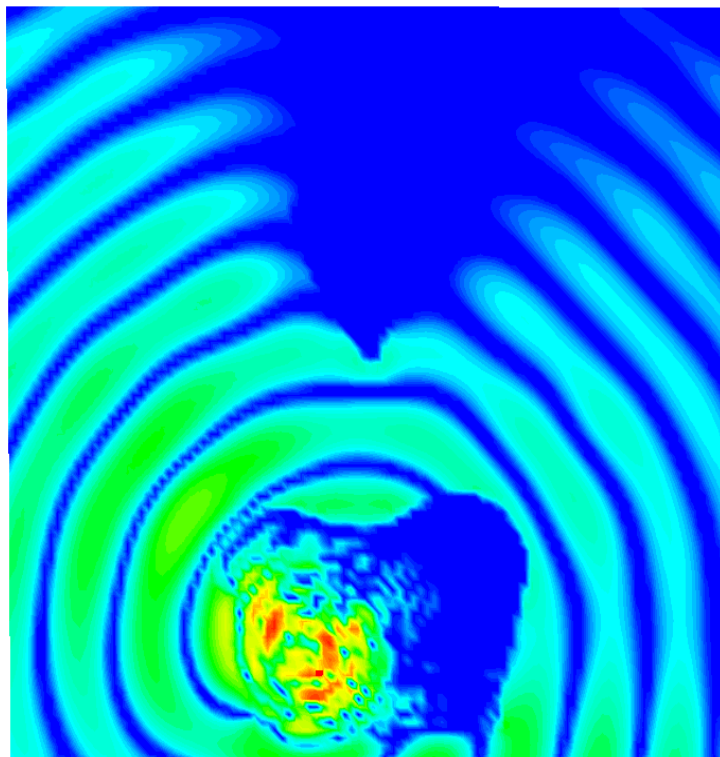
Iteration	Dimensions [mm]	Bandwidth [%]	VSWR in band
0	60 x 35	25.73	1.35
1	30 x 25	39.64	1.20
2	22 x 23	42.28	1.23



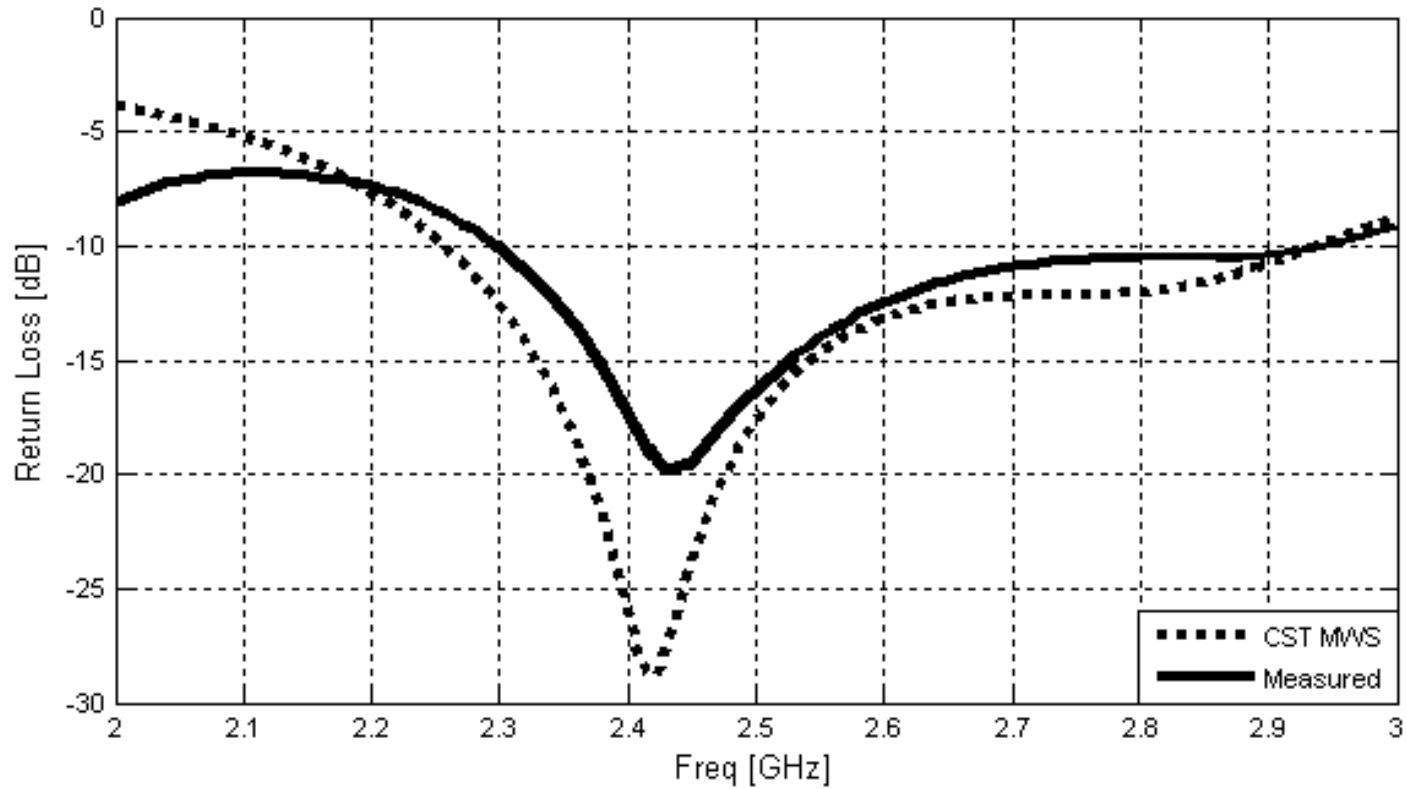
Fractal slot dipole (4)



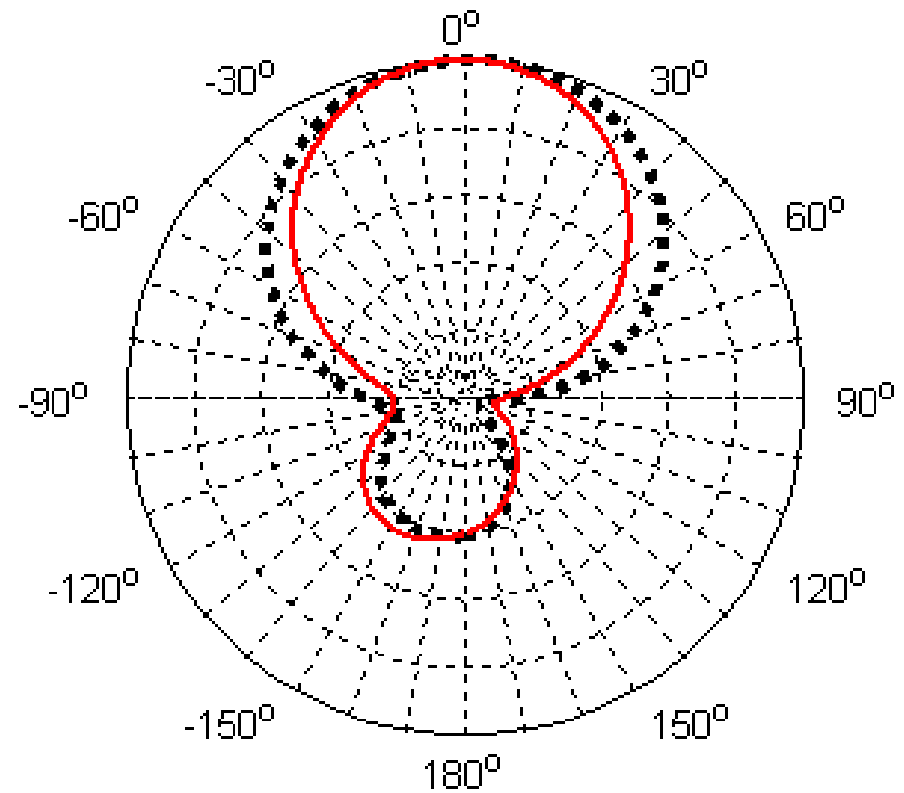
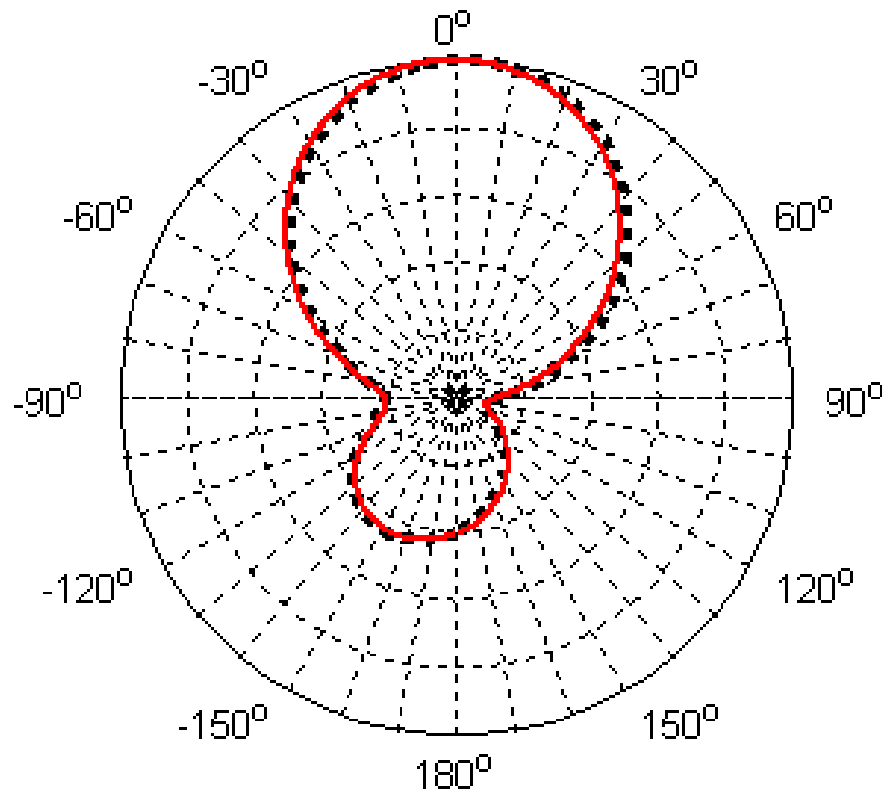
Fractal slot antenna (5)



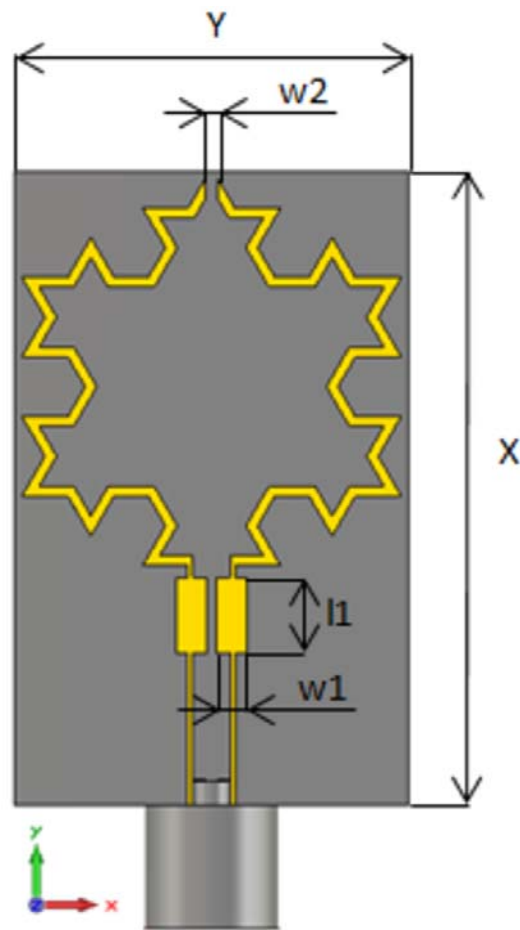
Fractal slot antenna (6)



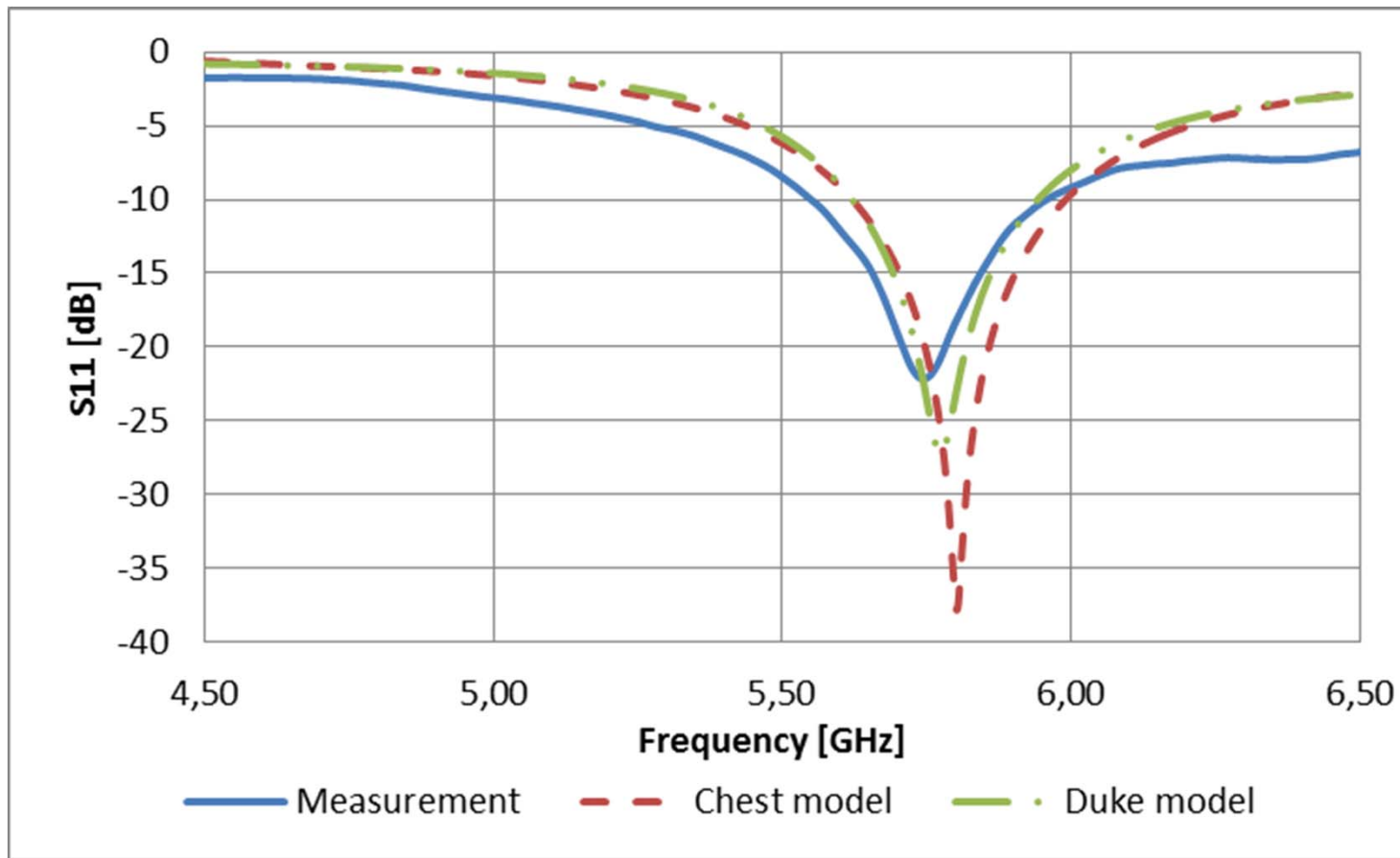
Fractal slot antenna (7)



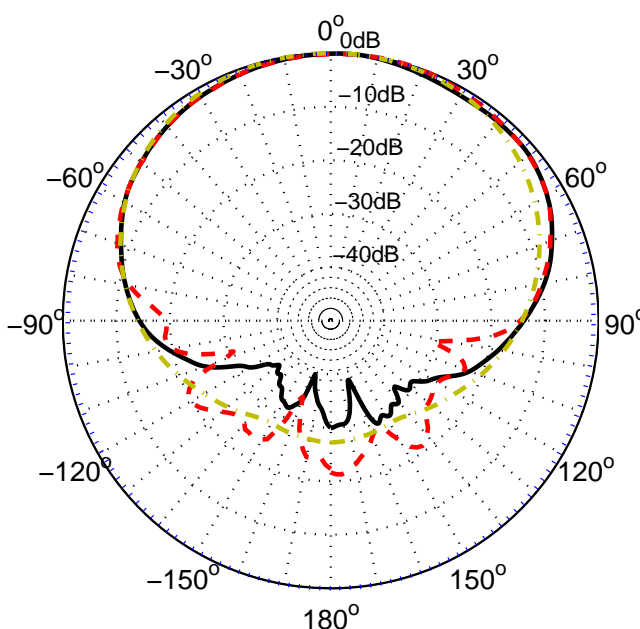
Koch loop antenna (1)



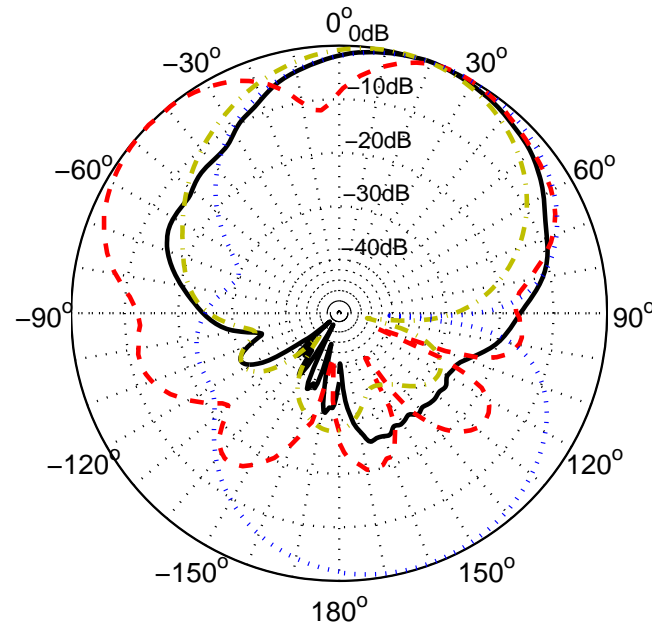
Koch loop antenna (2)



Koch loop antenna (3)



— Duke Model - - Measurement - - Chest Model



— Duke Model - - Measurement - - Chest Model Free Space

Slot loop antenna (1)

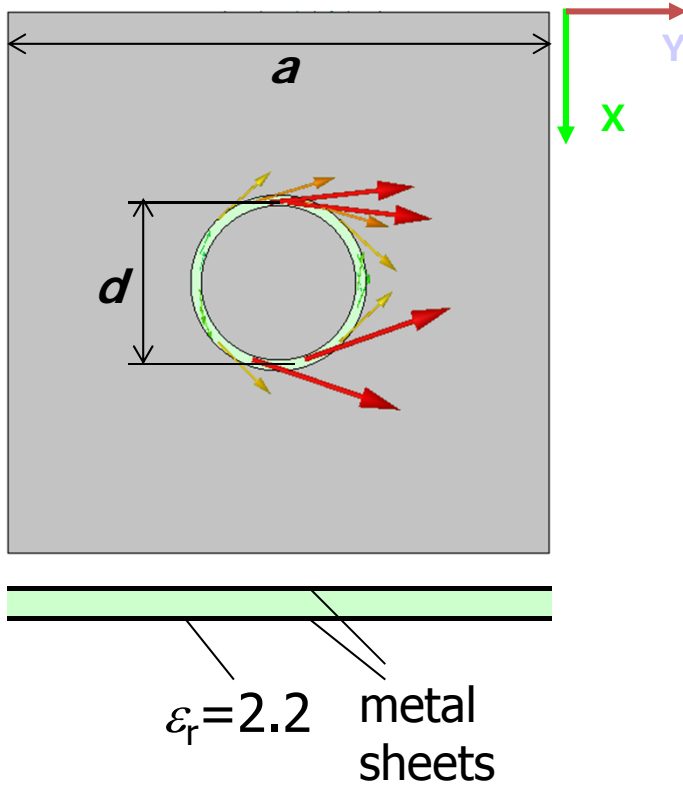
Parameters to be met:

- ❑ Frequency: 5.725 GHz to 5.875 GHz (ISM band)
- ❑ Input impedance: 50 Ω
- ❑ Radiation: omnidirectional in antenna plane
- ❑ Polarization: linear
- ❑ Impedance bandwidth: 2.6%
- ❑ Tested: on muscle-equivalent phantom

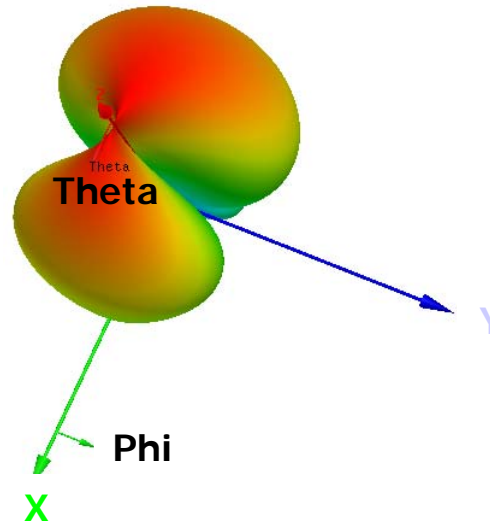
Slot loop antenna (2)

■ Circular loop antenna: first operating mode

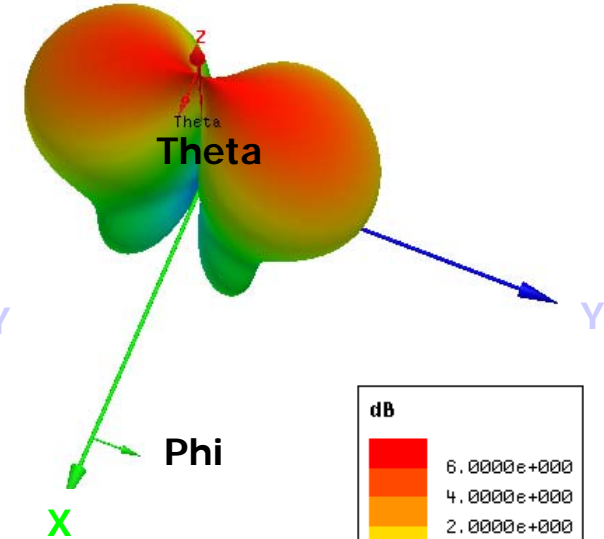
Equivalent magnetic current



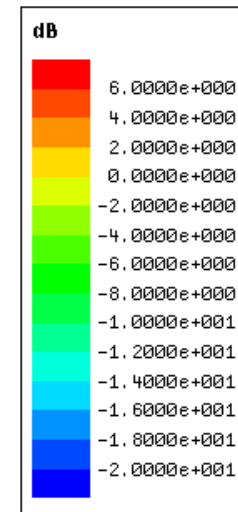
Gain Theta



Gain Phi

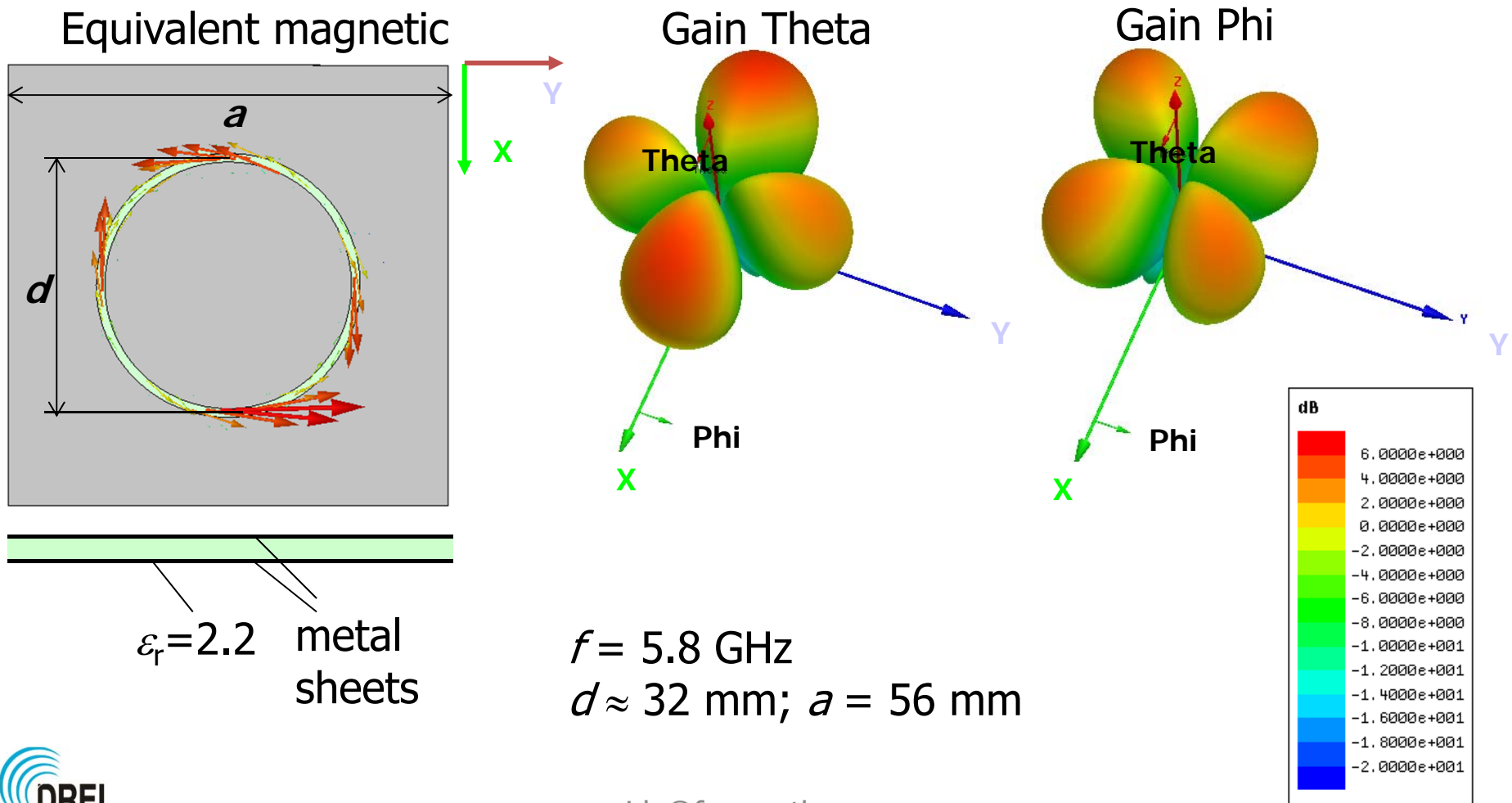


$f = 5.8 \text{ GHz}$
 $d \approx 16 \text{ mm}; a = 56 \text{ mm}$



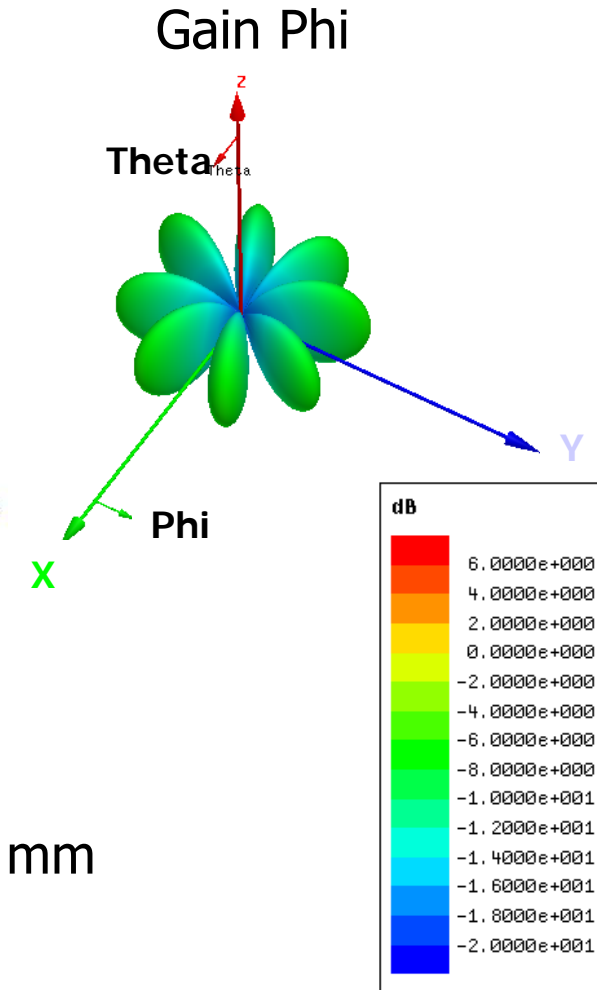
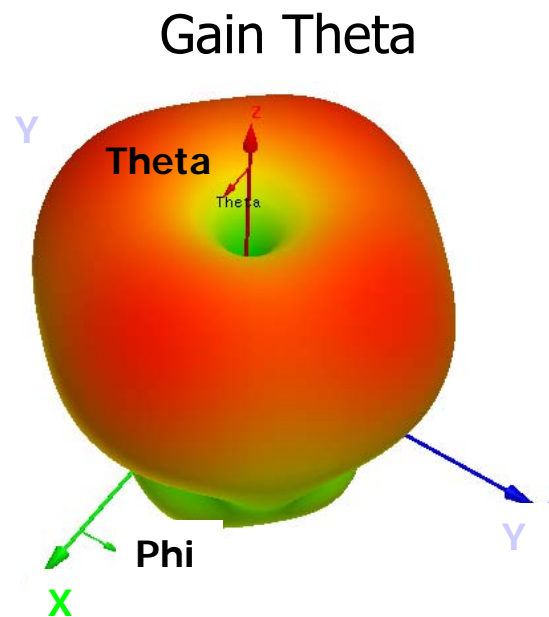
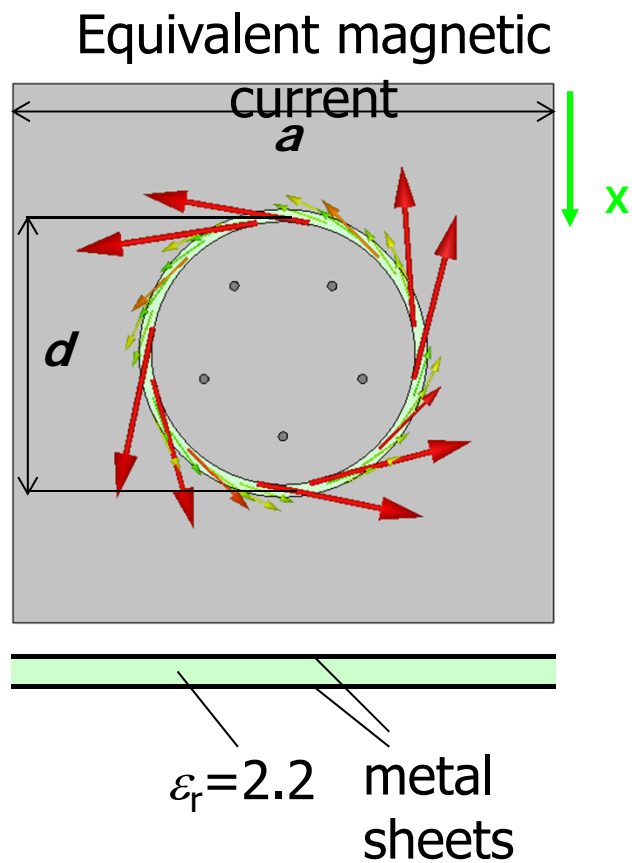
Slot loop antenna (3)

■ Circular loop antenna: second operating mode



Slot loop antenna (4)

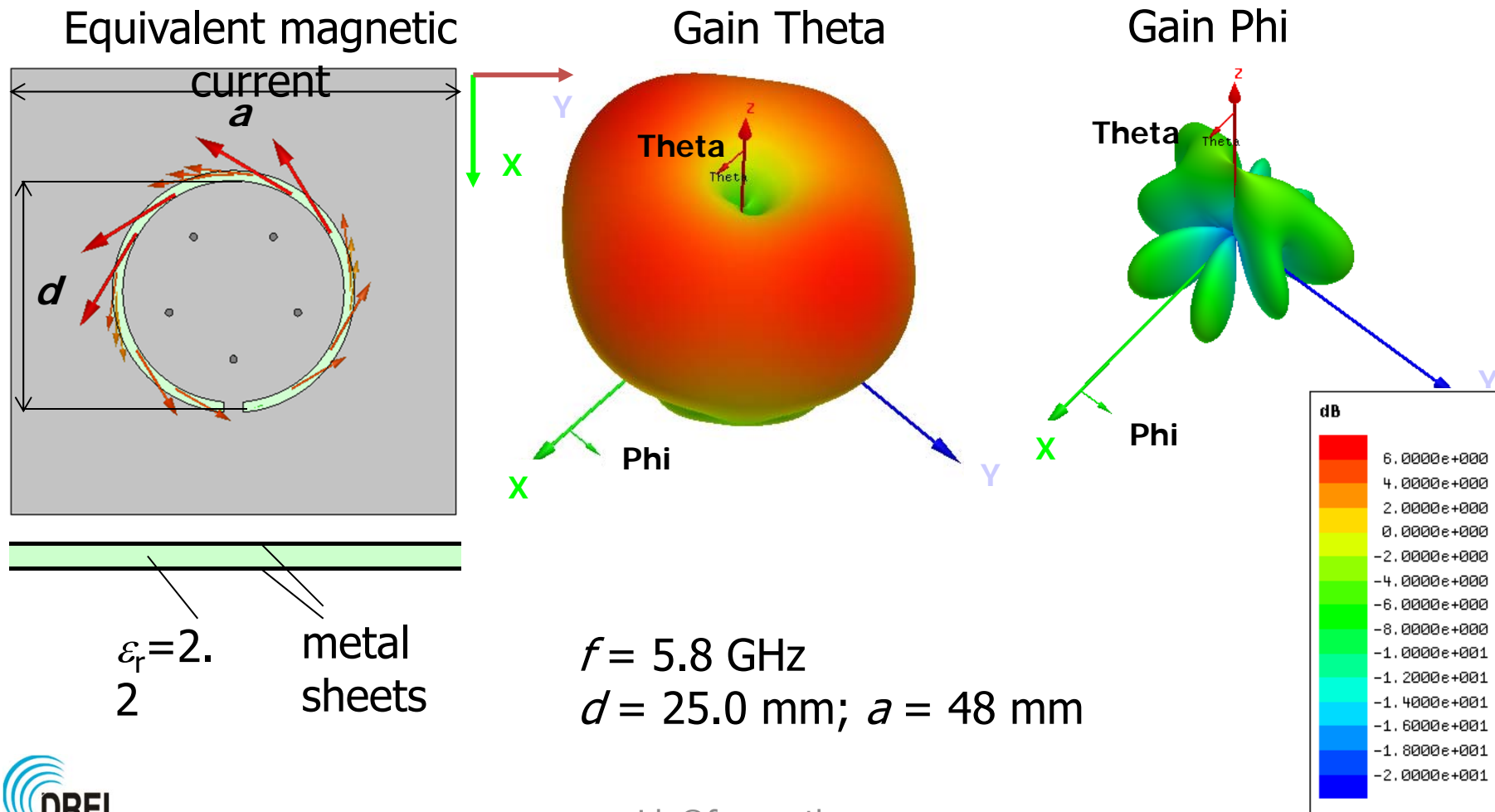
■ Circular loop antenna with 5 shorting vias



$f = 5.8 \text{ GHz}$
 $d = 24.5 \text{ mm}; a = 48 \text{ mm}$

Slot loop antenna (5)

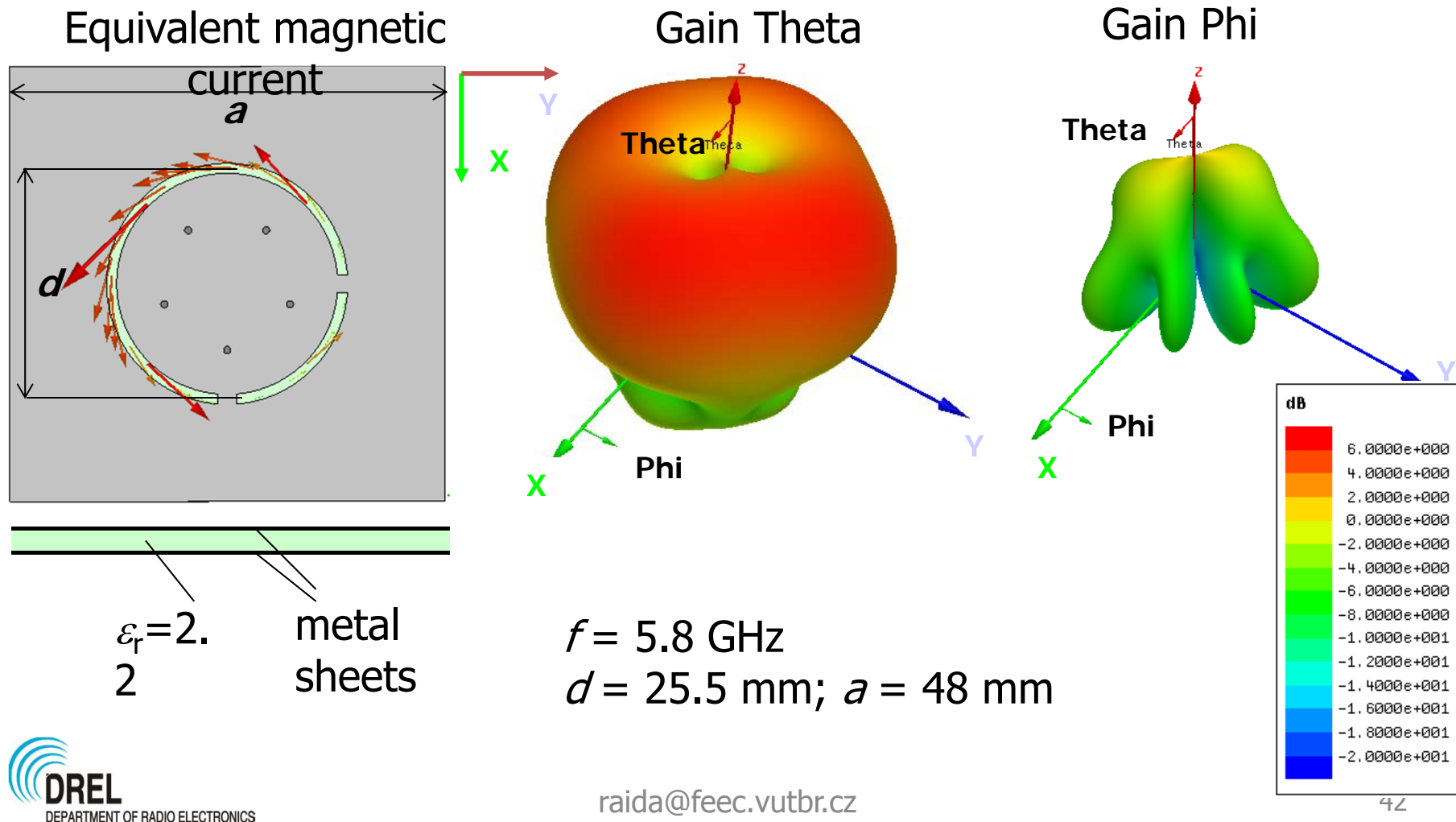
- Circular loop antenna with 5 shorting vias and 1 shorting strip



$f = 5.8 \text{ GHz}$
 $d = 25.0 \text{ mm}; a = 48 \text{ mm}$

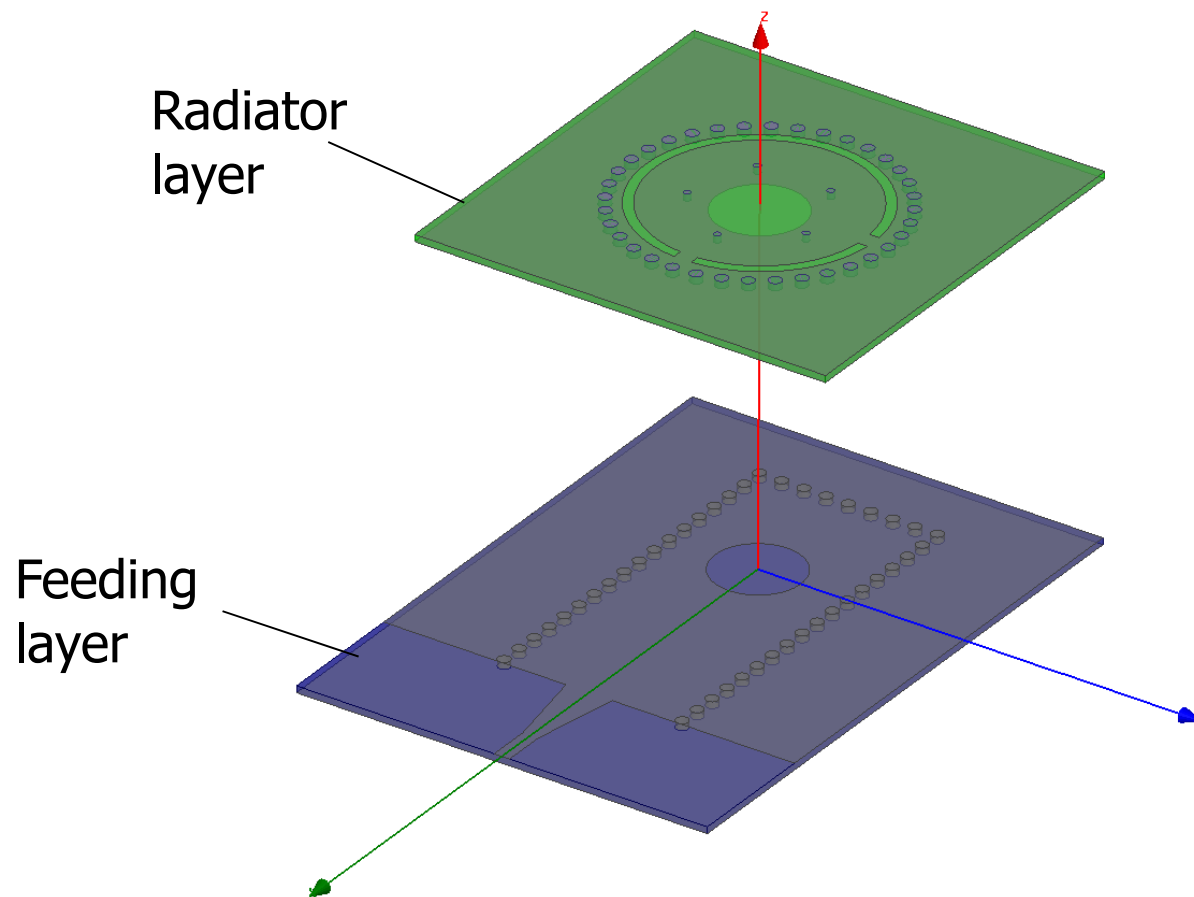
Slot loop antenna (6)

- Circular loop antenna with 5 shorting vias and 2 shorting strips



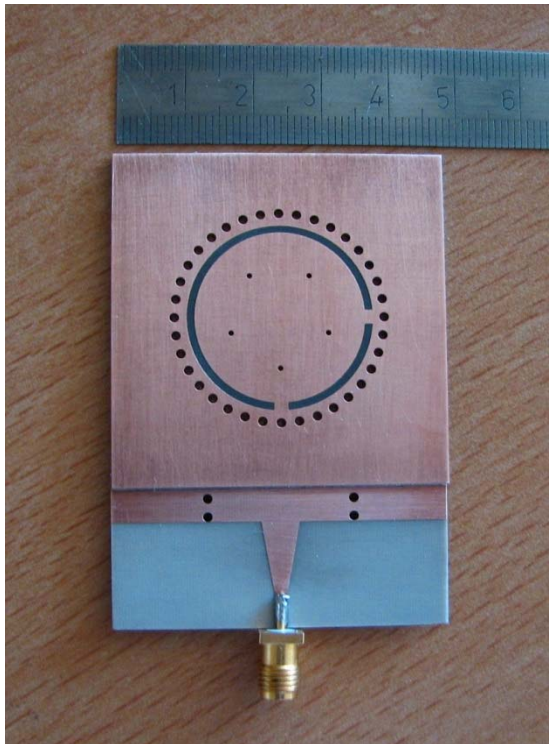
Slot loop antenna (7)

■ Antenna configuration

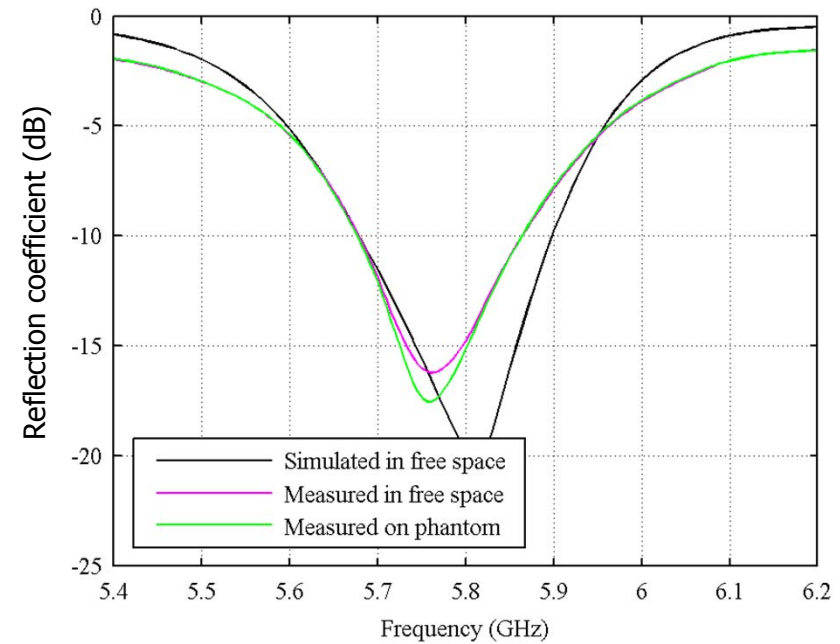


Slot loop antenna (8)

Fabricated antenna



Reflection coefficient



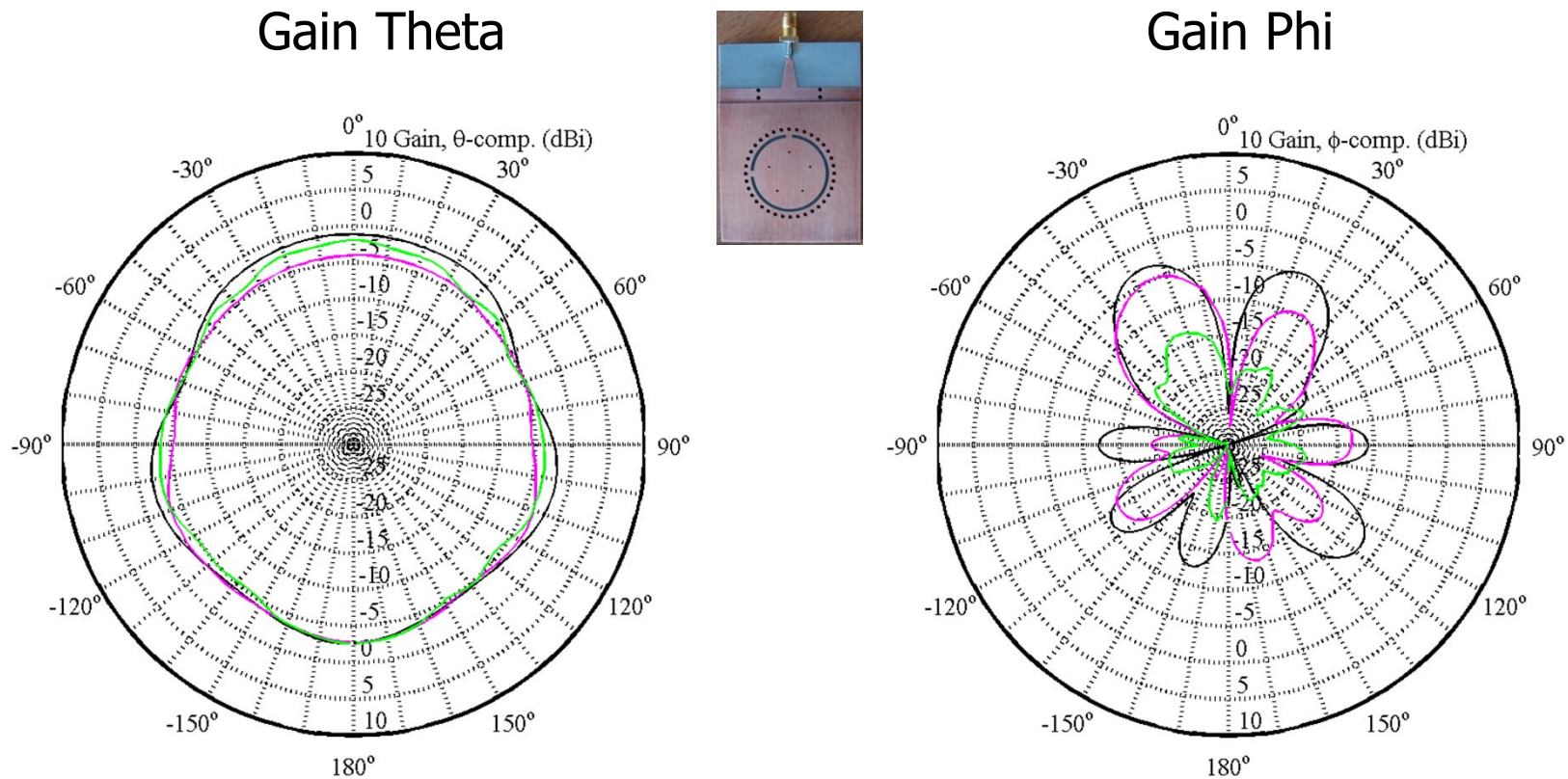
IEEE 802.11a frequency
band 5.725 – 5.825 GHz

raida@feec.vutbr.cz

ISM band 5.725 – 5.875 GHz

Slot loop antenna (9)

■ Comparing radiation in horizontal plane

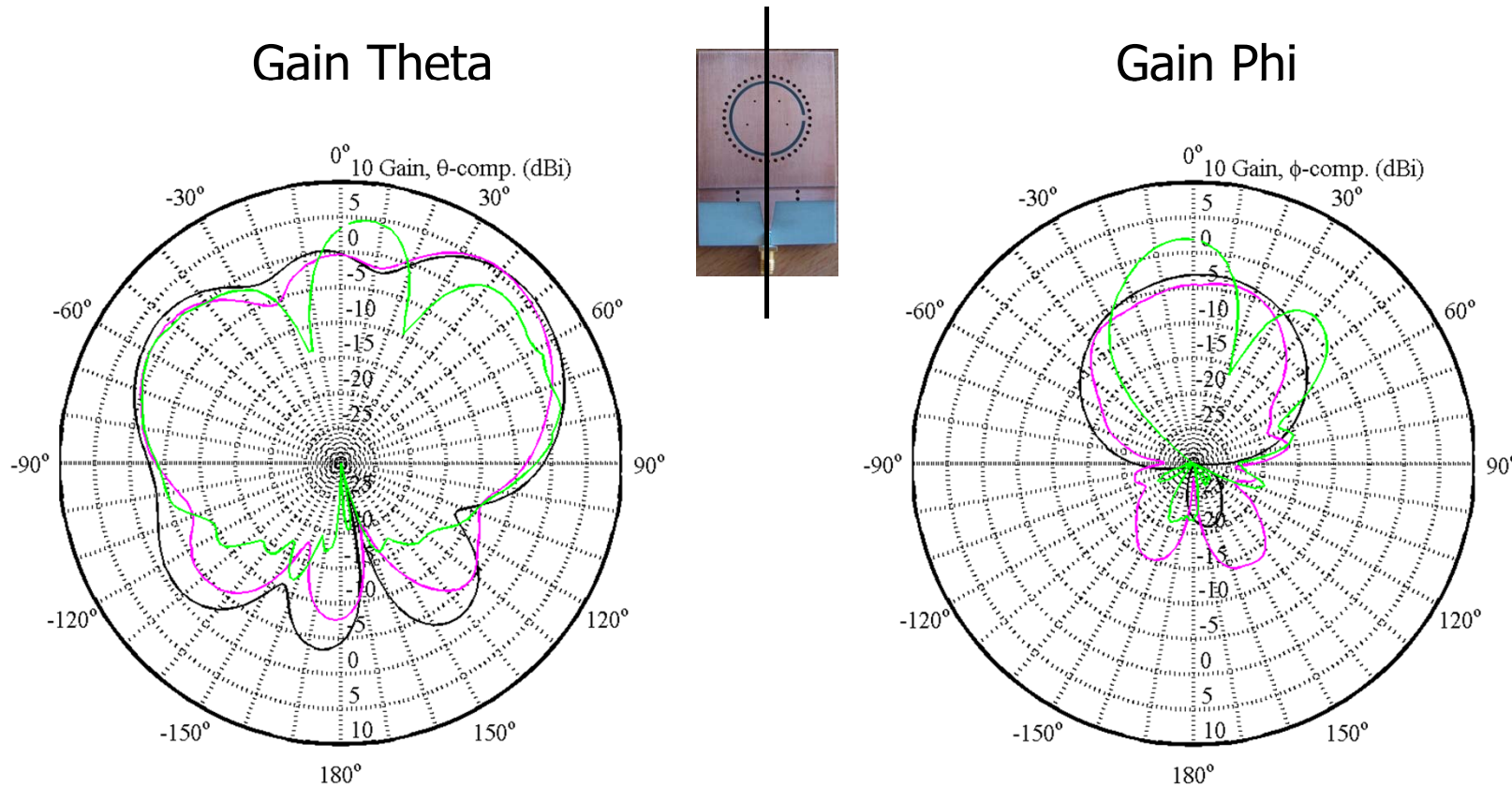


- Simulated in free space
- Measured in free space
- Measured on phantom

$f=5.8$ GHz

Slot loop antenna (10)

■ Comparing radiation in vertical plane



- Simulated in free space
- Measured in free space
- Measured on phantom

$f=5.8$ GHz

Summary

- Planar slot-line antennas suitable for wireless body area networks presented
- Slot dipoles: directional
- Slot loops: omnidirectional
- Testing
 - Good agreement of chest model and voxel phantom
 - Good agreement of simulation and measurement

Acknowledgements







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Research and Development for Innovation

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Thank you for your attention



raida@feec.vutbr.cz

Dept. Radio Electronics
Brno University of Technology
Technicka 12, 616 00 Brno, Czech Republic